

# Capacitance Meter Using ATmega128

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# Motivation

Capacitor tolerances are commonly  $\pm 20\%$ ... does this matter?

- ▶ Digital applications ✓
  - ▶ Usually just a decoupling cap, precise value doesn't matter
- ▶ Analog applications ✗
  - ▶ Used in filters, for which cutoff frequency is important
  - ▶ Matching requirements in data converters often  $< 0.1\%$
  - ▶ Exacerbated in high-frequency PLLs and RF transceivers

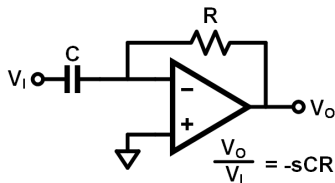
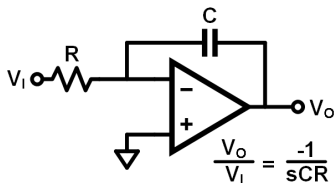


Figure 1: Examples of capacitors used in integrator (left) and differentiator (right)

# I-V Characteristics of Capacitor

Integral form of capacitor voltage:

$$V(t) = \frac{1}{C} \int_{t_0}^t I(\tau) d\tau + V(t_0)$$

$$C = \frac{1}{V(t) - V(t_0)} \int_{t_0}^t I(\tau) d\tau$$

Two assumptions to greatly simplify this: constant current  $I(\tau) = I_0$ , and initial condition  $V(t_0) = 0$ . Then:

$$C = \frac{1}{V(t)} \cdot I_0 \cdot (t - t_0)$$

$$C = \frac{I_0 \cdot \Delta t}{V(t)}$$

# I-V Characteristics of Capacitor

$$C = \frac{I_0 \cdot \Delta t}{V(t)}$$

## Two options to measure unknown C:

- 1) Measure  $V(t)$  after fixed  $\Delta t$ 
  - ▶ Some fixed delay using timer
  - ▶ Use analog-to-digital (ADC) to read  $V(t)$
  - ▶  $C \propto \frac{1}{V(t)}$  ❌

- 2) Count  $\Delta t$  to reach fixed  $V(t)$ 
  - ▶ Fixed threshold voltage with analog comparator
  - ▶ Count  $\Delta t$  with timer
  - ▶  $C \propto \Delta t$  ✅

# Analog Comparator within ATmega128

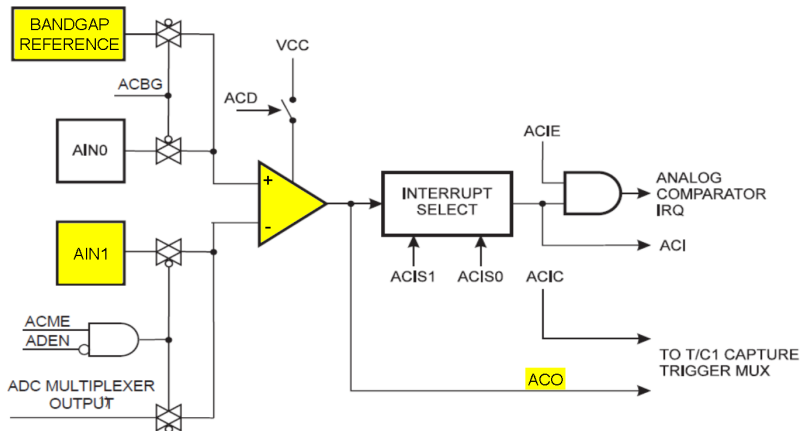
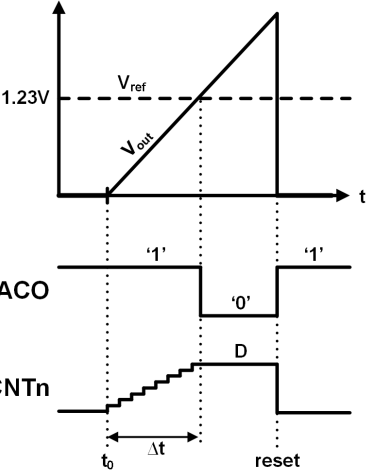
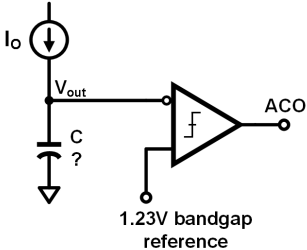


Figure 2: Analog Comparator block diagram<sup>1</sup>

Q: What is so great about having a bandgap reference?

<sup>1</sup>ATmega128 Datasheet (2011), p. 227

# Comparator Output Waveform



$$D = \frac{\Delta t}{T_{clk}} = \Delta t \cdot f_{clk}$$

Figure 3: Using analog comparator to track charging progress

# Implementing DC Current Source

Main points:

1. Feedback path forces  $V_+ = V_-$
2. Voltage drop over  $R_3$  determines output current
3. BJT acts as unity-gain current buffer
4. Works well for  $V_{out} < V_- - V_{ECsat}$ 
  - ▶  $R_2 = 10 \cdot R_1$  gives large headroom

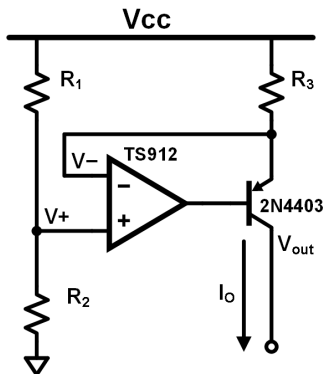


Figure 4: Providing constant current with opamp

$$\begin{aligned} I_o &= \frac{V_{CC} - V_-}{R_3} = \frac{V_{CC} - V_+}{R_3} \\ &= \frac{V_{CC} - \left(\frac{10}{11}\right)V_{CC}}{R_3} = \frac{V_{CC}}{11 \cdot R_3} \end{aligned}$$

# Determining D→C Transfer Function

From before:

$$C = \frac{I_O \cdot \Delta t}{V(t)}$$

where

$$\Delta t = \frac{D}{f_{clk}}$$

$$I_O = \frac{V_{cc}}{11 \cdot R_3}$$

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Then transfer function is given by:

$$C = \frac{V_{cc} \cdot D}{11 \cdot R_3 \cdot V_{ref} \cdot f_{clk}}$$



## Determining D→C Transfer Function

Assuming<sup>2</sup>  $V_{CC} = 5V$  and with  $R_3 = 1k$ :

$$\begin{aligned}C &= \frac{5 \cdot D}{11 \cdot 1k \cdot 1.23 \cdot 16M} = (2.30968 \cdot 10^{-11})D & [F] \\ &= 0.0230968 \cdot D & [nF] \\ 10 \cdot C &= 0.230968 \cdot D & [nF]\end{aligned}$$

- ▶ For digital arithmetic, calculating  $10C$  makes result “look like” floating point number
  - ▶ Just insert decimal point before writing to LCD
- ▶ D comes from 16-bit TCNTn, so range is 23pF to  $1.5\mu F$

Q: What are two ways to change this range?

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<sup>2</sup>Those of you using USB power may have different values

# Final Hardware Design

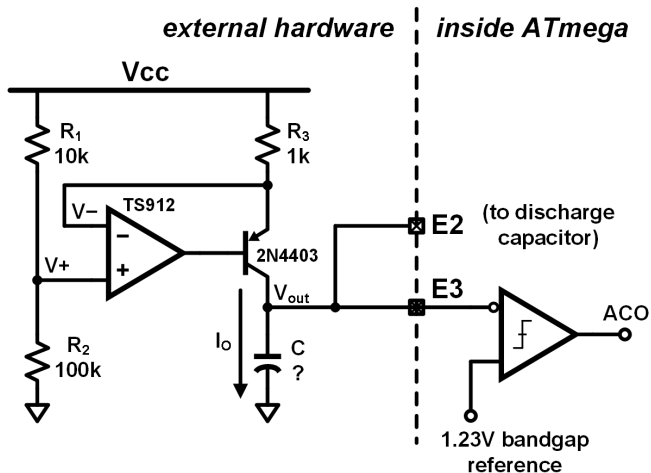
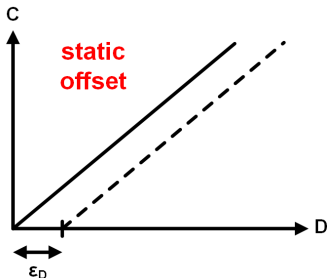


Figure 5: Schematic of capacitance meter with ATmega128

# Sources of Error

- ▶ Offset errors
  - ▶ Capacitor doesn't fully discharge
  - ▶ Counter starts before capacitor starts charging
  - ▶ Equivalent series resistance (ESR)
  - ▶ I/O pin contributes some fixed capacitance
- ▶ Gain errors
  - ▶ Supply voltage variations
  - ▶ Component tolerances
- ▶ Nonlinearity
  - ▶ Current source nonlinearity
  - ▶ Capacitance may change with voltage

# Sources of Error



Which of these errors are easy to correct?

Which are difficult to correct?

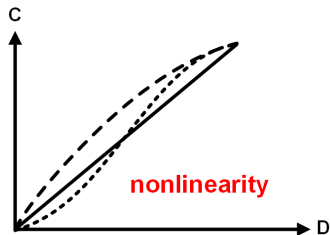


Figure 6: Illustration of  $D \rightarrow C$  errors

# Calibrating Errors

$$C = m \cdot (D - \epsilon_D)$$

- ▶ Static offset:  $\epsilon_D$  is count when  $C = 0$
- ▶ Gain error: nominally 0.230968, but can be determined experimentally using well-known capacitor values
- ▶ Nonlinearity: sometimes implemented as piece-wise gain error, or as a look-up table
  - ▶ Generally a more complicated beast, outside the scope of this discussion

## Relevance to ADC Design

Alternatively, suppose that  $C$  is known and reference voltage is unknown  $V_{IN}$ :

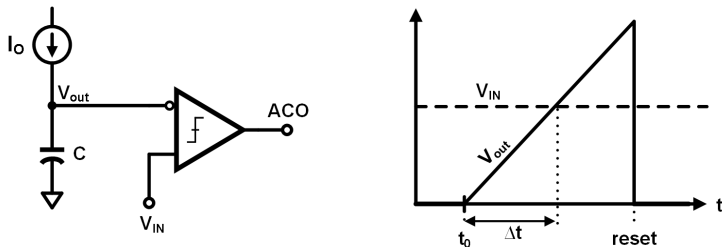


Figure 7: Capacitance meter reconfigured as ADC

ADC transfer function looks like:

$$V_{IN} = \frac{I_o \cdot D}{C \cdot f_{clk}}$$

## Relevance to ADC Design

Integrating quantizer is one of many ADC architectures. Others include:

- ▶ Flash
- ▶ Pipeline
- ▶ Successive Approximation
- ▶ Incremental
- ▶  $\Delta\Sigma$  modulator

ADCs are *inherently* mixed-signal design... requires background in both analog (ECE 422/423/520) and digital (ECE 46x/471/474) circuits, as well as scripting (MATLAB, Bash, HSpice).