

SECTION 1: DIODES

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Course Overview

Course Overview

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- Your previous electrical engineering courses focused on circuits containing:
 - ▣ Sources
 - Voltage and current
 - Independent and dependent
 - ▣ Passive devices
 - Resistors, capacitors, inductors
 - ▣ Opamps
 - Integrated circuits (ICs)
 - Semiconductor devices
 - Comprised of transistors

- In this course, our focus will be on ***semiconductor devices***:
 - ▣ Diodes
 - ▣ Transistors

Course Overview

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- We will learn to ***analyze and design circuits using semiconductor devices*** – diodes and transistors
- Our primary focus will be ***discrete, analog circuits***
- ***Discrete*** circuits:
 - Built using individual discrete components (transistors, diodes, Rs, Ls, Cs)
 - Not integrated circuits (ICs), though we will be laying the foundation for IC design
- ***Analog*** circuits:
 - Voltages/currents can vary ***continuously***
 - As opposed to digital circuits, where quantities can only assume discrete values

Course Overview

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- **Section 1: Diodes**
 - Introduction to semiconductors
 - Rectifier circuits
- **Section 2: Bipolar Junction Transistors**
 - Device characteristics and models
- **Section 3: BJT Amplifiers**
 - Various amplifier circuits using BJTs
- **Section 4: MOSFETs**
 - Device characteristics and models
- **Section 5: MOSFET Amplifiers**
 - Various amplifier circuits using MOSFETs
- **Section 6: Integrated Circuit Building Blocks**
 - Intro to basic components of integrated circuits

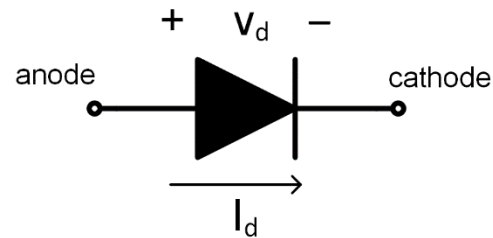
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Diode Fundamentals

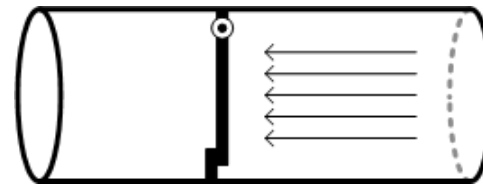
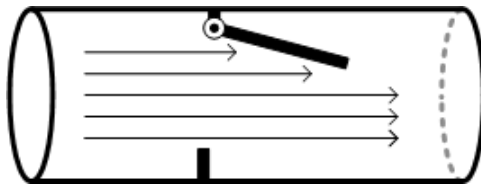
Diodes

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- Two terminal electrical components that ***allow current to flow in one direction only***



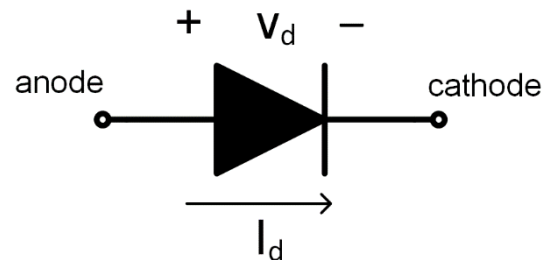
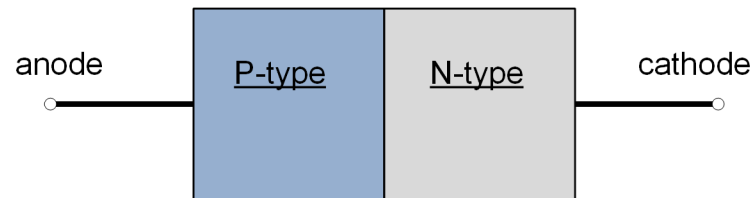
- Current can flow from ***anode to cathode***, but not from cathode to anode
- Analogous to ***check valves***
 - Fluid can flow in one direction only:



Diodes – PN Junctions

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- Diodes are ***semiconductor*** devices
 - ▣ Formed from the junction of two dissimilar semiconductor materials
 - ***P-type*** and ***N-type*** semiconductors
 - ▣ ***P-N junctions***:



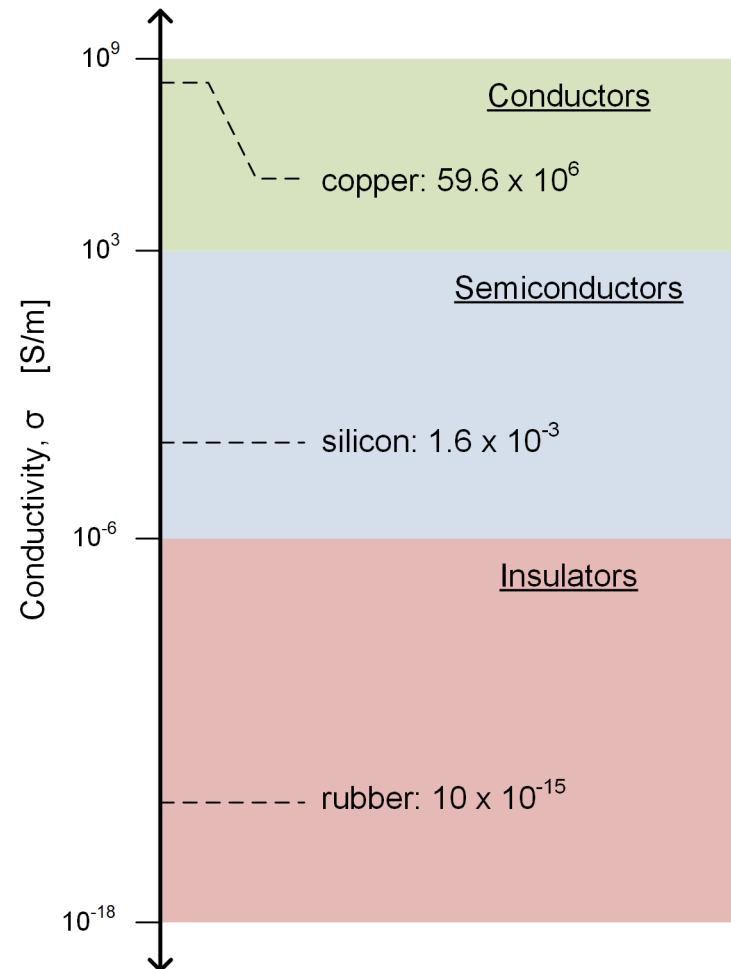
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Semiconductor Fundamentals

Semiconductors

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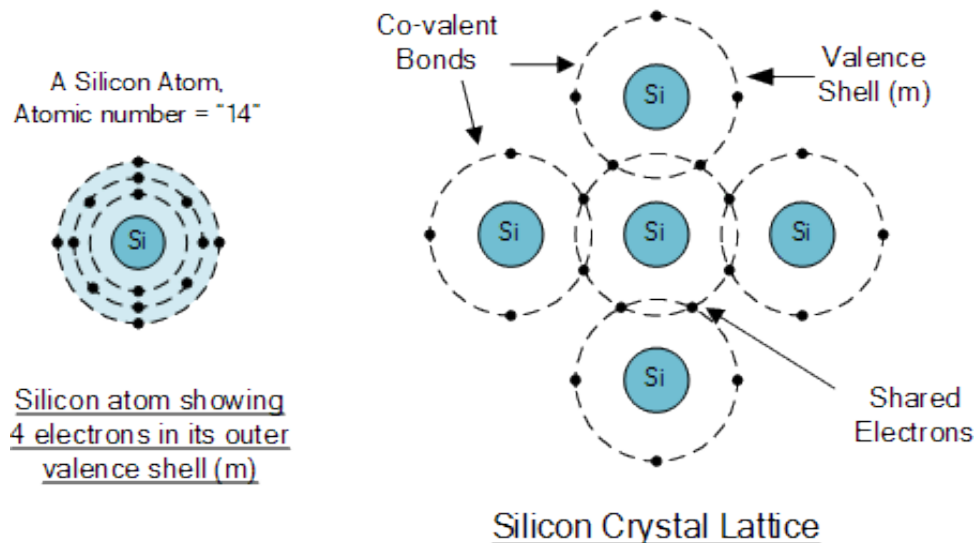
- Semiconductors have ***conductivities*** between those of conductors and insulators
- Conductivity can be modified by adding impurities
 - ▣ ***Doping***



Semiconductors

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- Semiconductors have ***four valence electrons***
 - ▣ Valence bands are half full
 - ▣ Atoms bond together in a crystalline lattice structure
 - ▣ Each atom shares covalent bonds with four adjacent atoms
 - ▣ Few free charge carriers – poor conductivity



Semiconductors – Doping

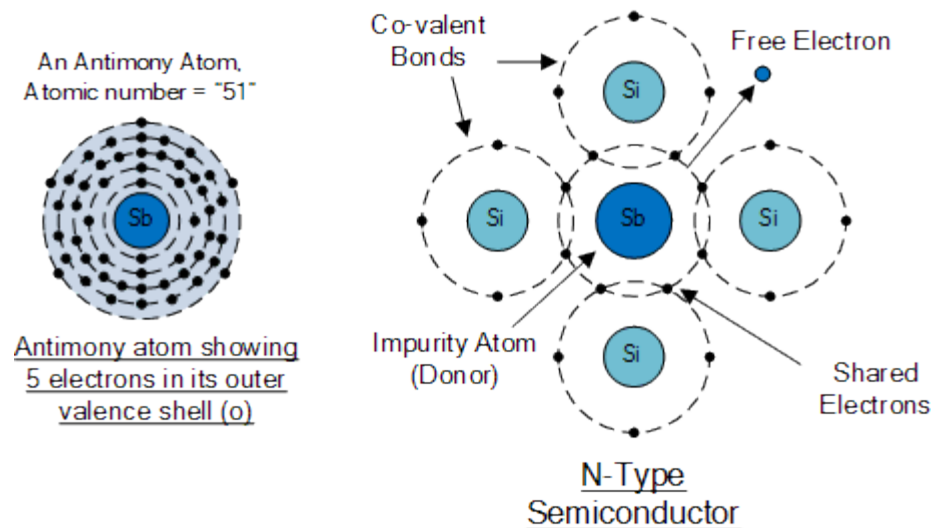
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- ***Intrinsic*** semiconductors are poor conductors
- Conductivity can be altered by adding ***impurities*** through a process called ***doping***
- ***Doping***: addition of a small amount of another element to the intrinsic semiconductor
 - ▣ Low concentration: 10 ppb – 100 ppm
 - ▣ Free charge carriers are increased
 - ▣ Conductivity is increased
- Type of dopant determines type of semiconductor
 - ▣ N-type or P-type

Semiconductors – N-Type

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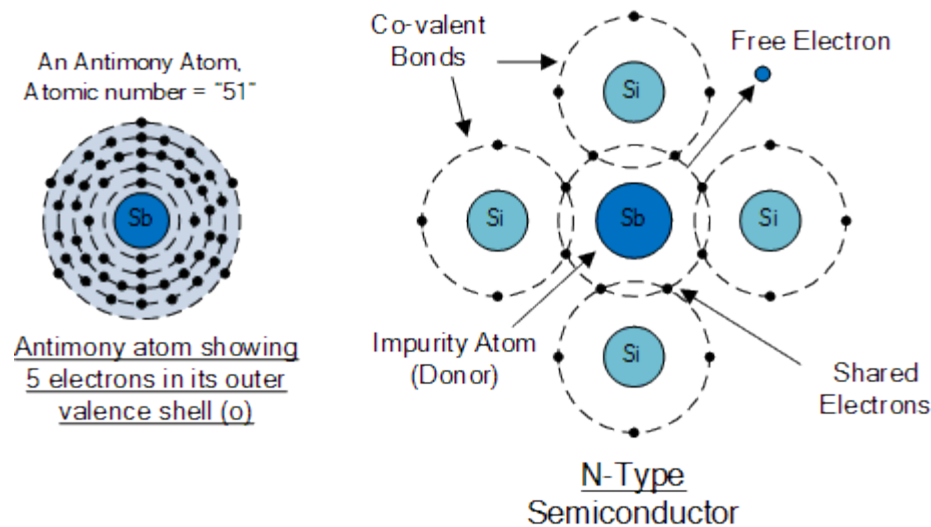
- **N-type** material created by doping with **pentavalent** dopant atoms
 - Five valence electrons
 - E.g., phosphorous (P) or antimony (Sb)
 - Bonding with four Si atoms creates a **free electron**
 - Dopants are called **donors** – they donate free electrons



Semiconductors – N-Type

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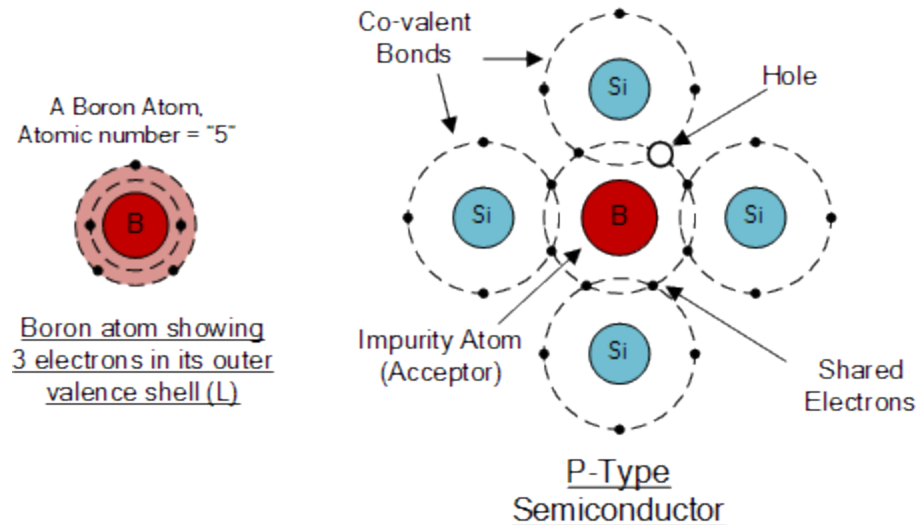
- Donor atoms lose an electron
 - ▣ These free electrons are **majority carriers**
- There will be some holes as well
 - ▣ Thermally-generated
 - ▣ Much lower concentration – **minority carriers**



Semiconductors – P-Type

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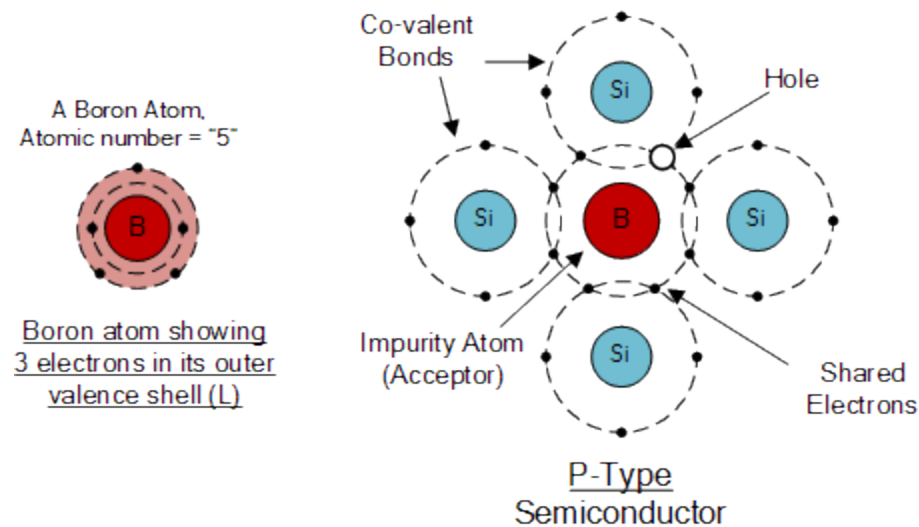
- **P-type** material created by doping with **trivalent** dopant atoms
 - Three valence electrons
 - E.g., Boron (B)
 - Bonding with four Si atoms creates a **hole**
 - Dopants are called **acceptors** – they accept an electron from a Si atom



Semiconductors – P-Type

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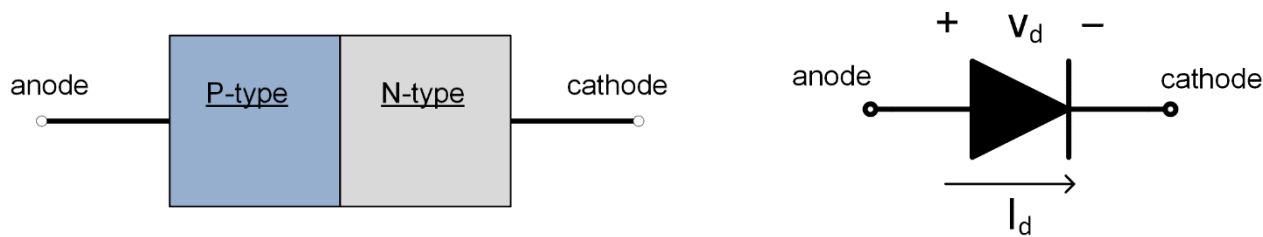
- Acceptor atoms introduce a hole
 - ▣ These holes are **majority carriers**
- There will be some free electrons as well
 - ▣ Thermally-generated
 - ▣ Much lower concentration – **minority carriers**



P-N Junctions

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- P-N junctions – diodes – are formed by joining P-type and N-type semiconductors

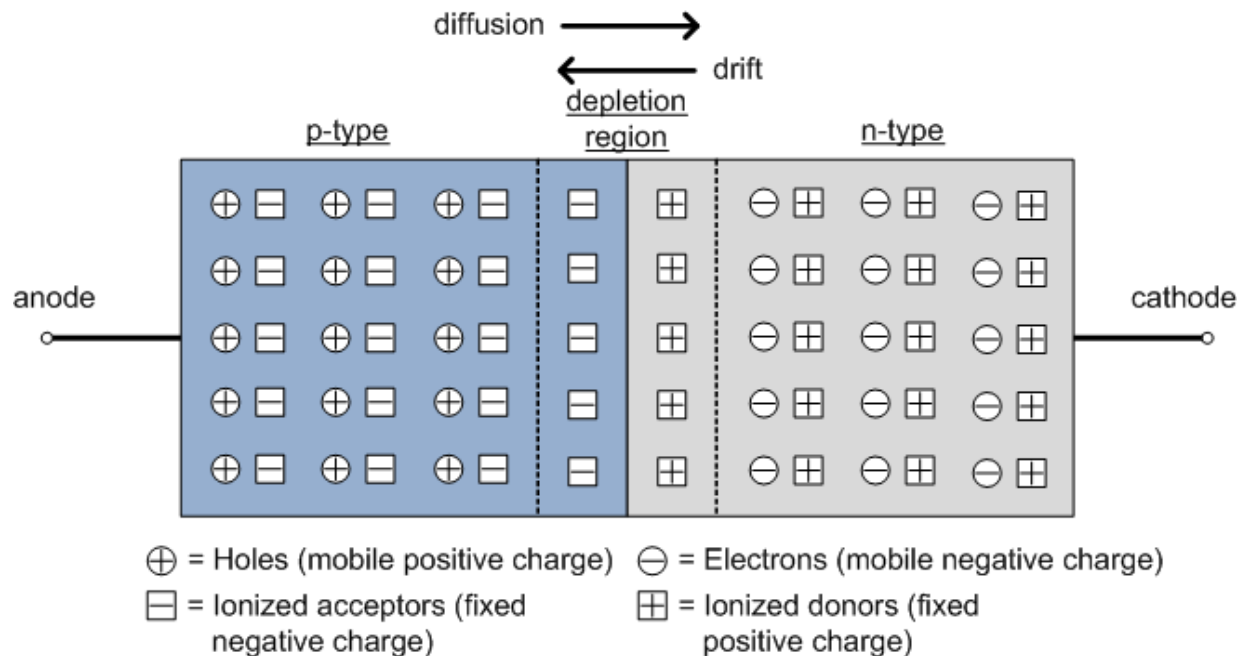


- P-type on one side of the junction
 - ▣ Majority carriers: **holes** – mobile positive charge
- N-type on the other
 - ▣ Majority carriers: **electrons** – mobile negative charge
- At the junction:
 - ▣ Holes want to **diffuse** across into the N-type material
 - ▣ Electrons want to **diffuse** across into the P-type material

P-N Junctions

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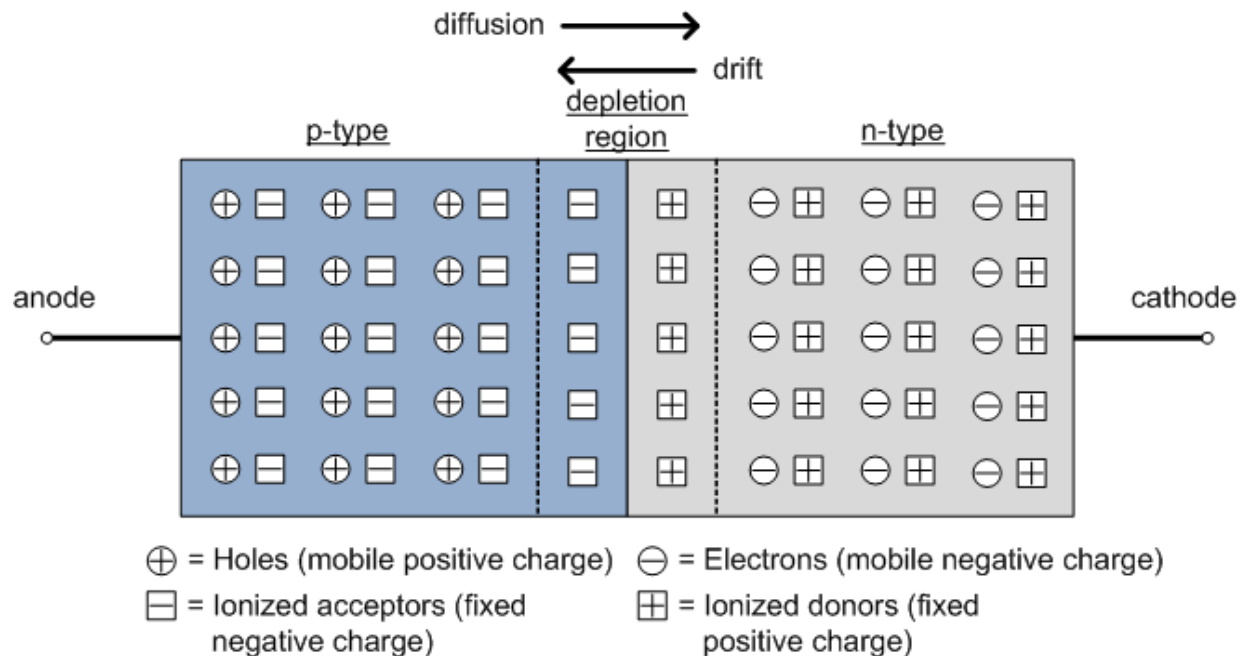
- Mobile holes want to diffuse from p-type to n-type
- Mobile electrons want to diffuse from n-type to p-type
- ***Space-charge layer*** or ***depletion region*** is created
- Diffusion drives negative charge one way, positive charge the other



P-N Junctions

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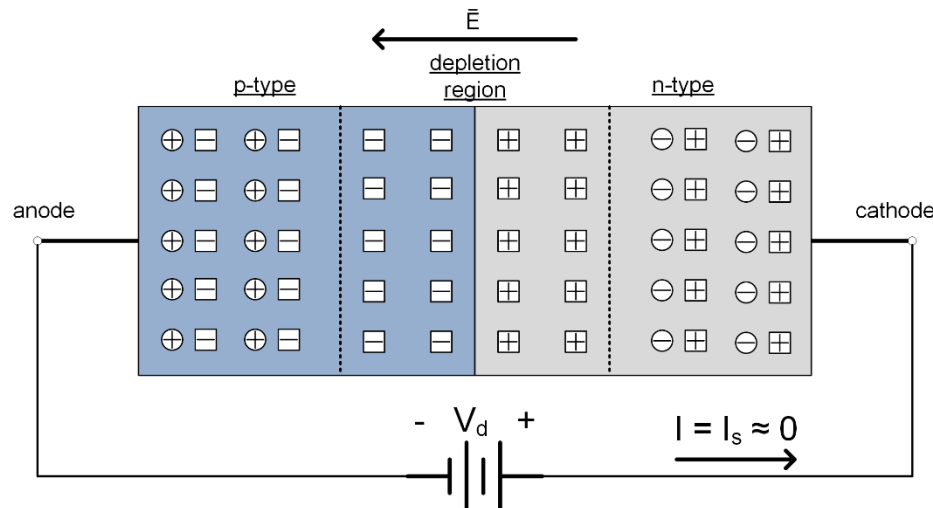
- Diffusion causes a **charge gradient** across the junction
 - ▣ **Electric field** established
- E-field drives mobile charge back in the opposite direction
 - ▣ **Drift** opposes and limits diffusion



P-N Junction – Reverse Bias

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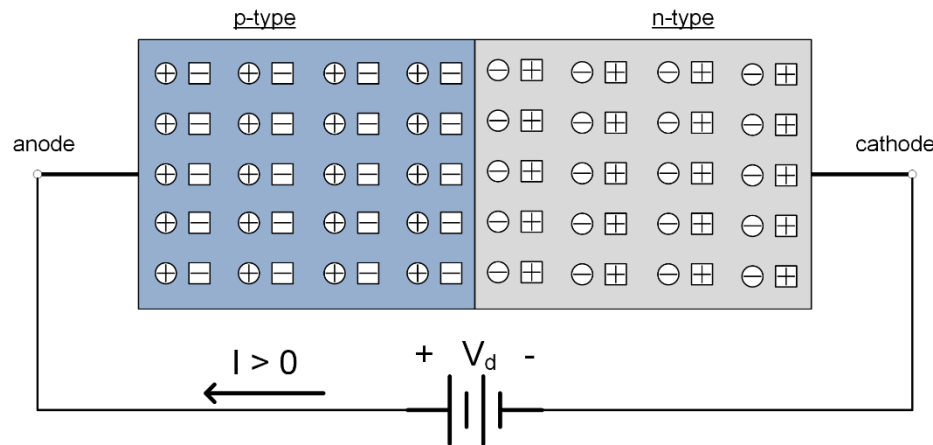
- Reverse-bias voltage applied – $V_d < 0$ V
 - ▣ Applied voltage adds to depletion region E-field
 - ▣ Additional diffusion to balance larger E-field
 - ▣ Depletion region expands
 - ▣ Negligible reverse current flow due to **drift** of thermally-generated **minority carriers**



P-N Junction – Forward Bias

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- Forward-bias voltage applied – $V_d > 0$ V
 - ▣ Applied voltage reduces depletion region E-field
 - ▣ Depletion region shrinks
- Significant forward current flows
 - ▣ **Majority carriers diffuse** across the junction
 - ▣ Carried to diode terminals by applied E-field



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Diode I-V Characteristics

Diode I-V Characteristics

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□ Three operating regions:

■ Forward biased

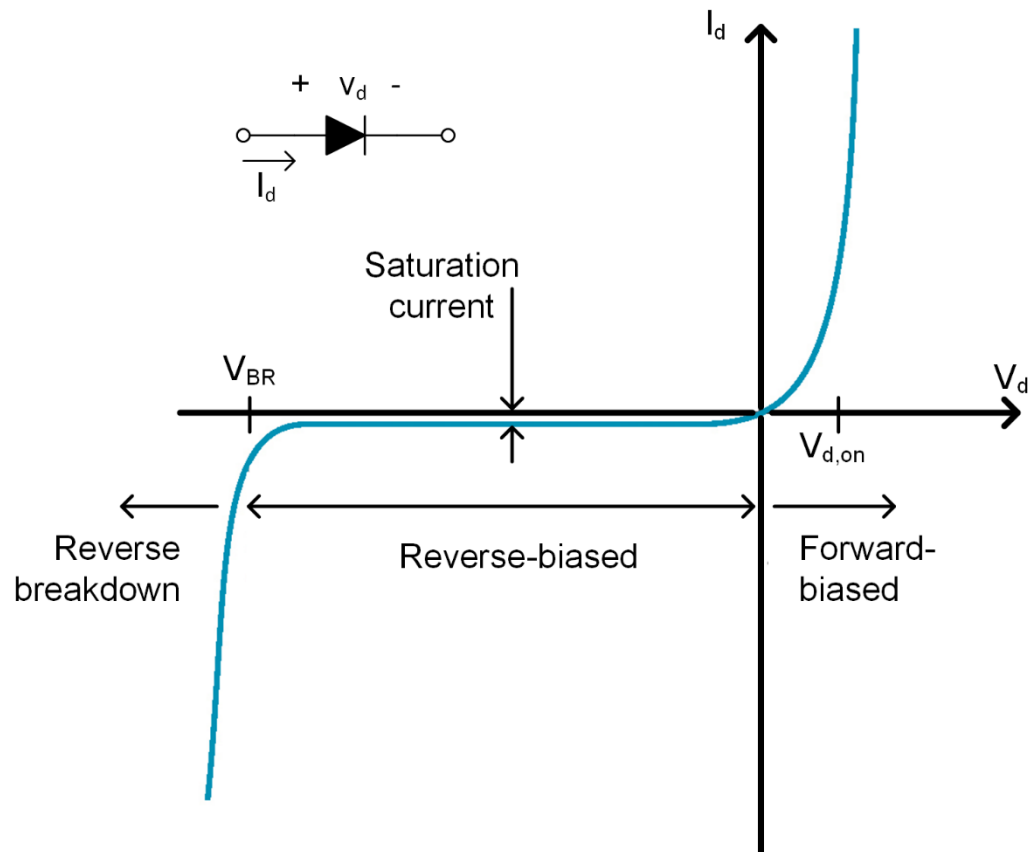
- $V_d > 0$ V
- Current flows from anode to cathode

■ Reverse biased

- $V_d < 0$ V
- Negligible, nearly-constant saturation current, I_s , flow from cathode to anode

■ Reverse breakdown

- $V_d < V_{BR}$
- Large reverse current flows



Shockley Equation

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- In the forward- and reverse-biased regions, diode behavior can be approximated by the **Shockley Equation**:

$$I_d = I_s \left(e^{\frac{V_d}{V_{th}}} - 1 \right)$$

- V_{th} is the thermal voltage

$$V_{th} = \frac{kT}{q}$$

- $k = 1.38 \times 10^{-23} \text{ J/K}$ is **Boltzmann's constant**
- $q = 1.6 \times 10^{-19} \text{ C}$ is the charge of an electron
- At $T = 300 \text{ K}$, $V_{th} \approx 26 \text{ mV}$
- I_s is the saturation current
 - Typically very small, e.g., $I_s \approx 35 \text{ pA}$

Shockley Equation

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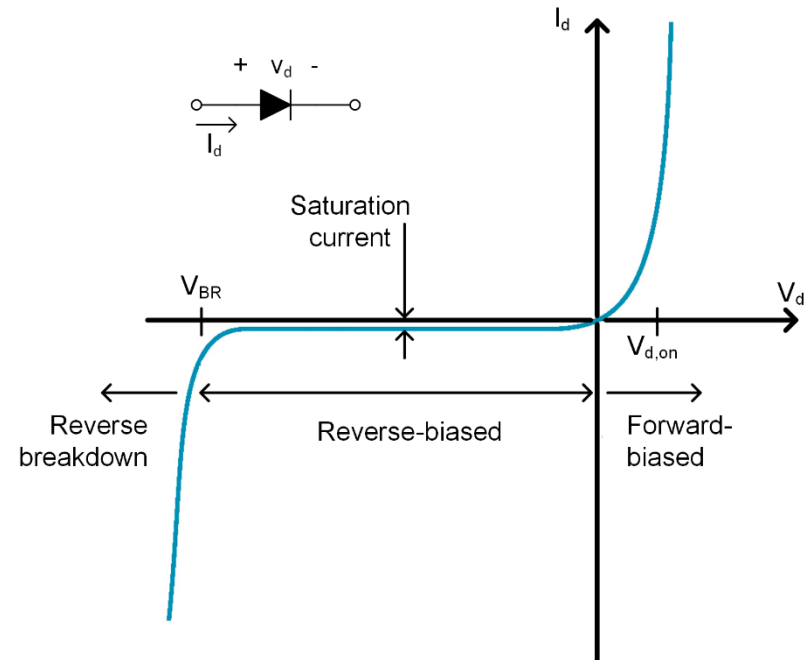
$$I_d = I_S \left(e^{\frac{V_d}{V_{th}}} - 1 \right)$$

- In forward bias, the $-I_S$ term is negligible
 - ▣ Exponential current-voltage relationship

$$I_d \approx I_S e^{\frac{V_d}{V_{th}}}$$

- In reverse bias, the exponential term is negligible
 - ▣ Nearly-constant, and small, reverse current

$$I_d \approx -I_S$$



Reverse Breakdown

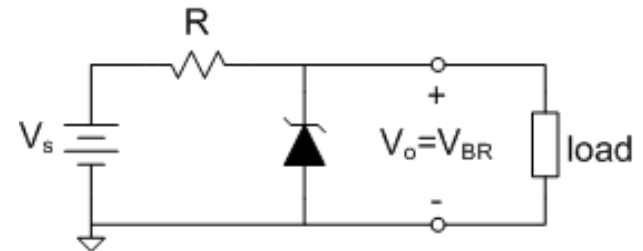
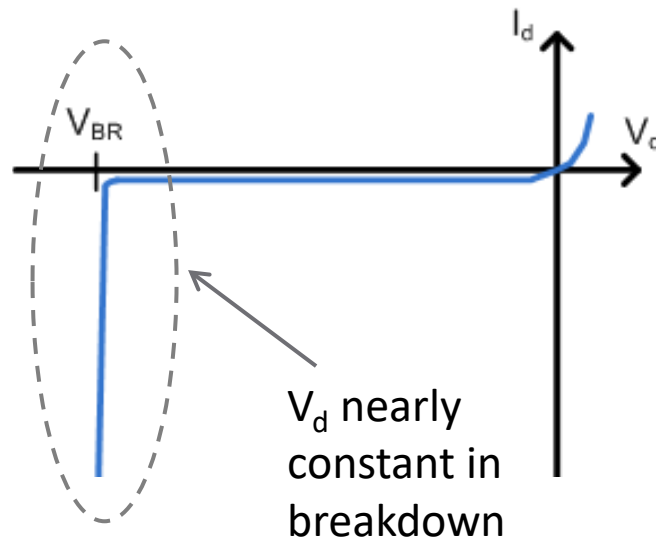
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- Shockley equation does not describe reverse-breakdown behavior
- Typical Breakdown voltages are in the range of a few volts to hundreds of volts
- Typically, we want to avoid exceeding the breakdown voltage
- However, one class of diodes is designed to be used in breakdown: ***Zener diodes***

Zener Diodes

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- **Zener diodes** designed to have a very steep I-V characteristic in the breakdown region
 - V_d is almost constant, independent of current
 - Useful in voltage regulation or voltage reference circuits
 - Schematic symbol:



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Diode Models

Diode Models

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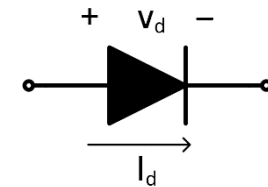
- Need a ***diode model*** to enable circuit analysis
- Shockley equation is one model
 - Simplified, but still complex for hand analysis
- Can trade off complexity and accuracy
 - Choose the simplest possible model that provides acceptable accuracy
- We'll look at three much simpler models
 - Appropriate for first-order type of analyses
 - Ideal diode model
 - Nearly-ideal diode model
 - Nearly-ideal model with resistance

Ideal Diode Model

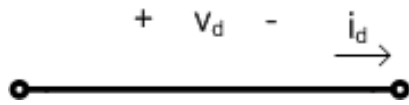
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□ ***Ideal diode model:***

- Short circuit for forward-bias current
- Open circuit for reverse-bias voltage

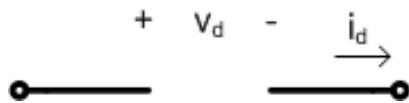


□ ***Forward-bias equivalent circuit:***

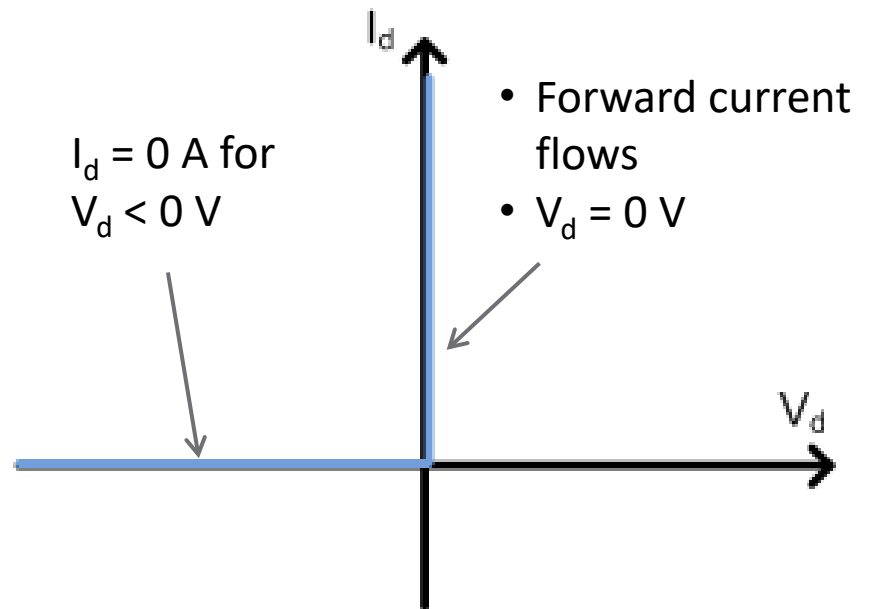


- $V_d = 0 \text{ V}$
- $I_d > 0 \text{ A}$

□ ***Reverse-bias circuit:***



- $V_d < 0 \text{ V}$
- $I_d = 0 \text{ A}$

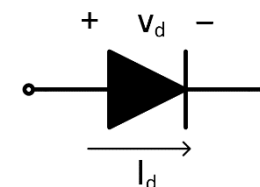


Nearly-Ideal Diode Model

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□ **Nearly-ideal diode model:**

- Accounts for diode forward voltage
- Open circuit for $V_d < V_{d,on}$

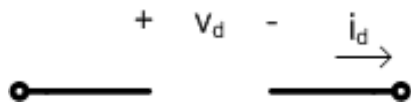


□ **Forward-bias equivalent circuit:**

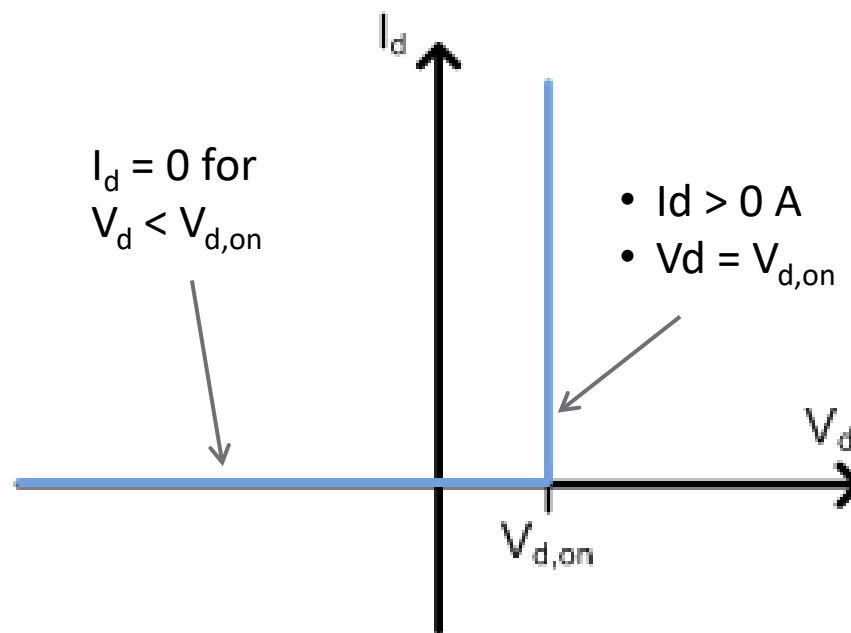


- $V_d = V_{d,on} \approx 700 \text{ mV}$
- $I_d > 0 \text{ A}$

□ **Reverse-bias circuit:**



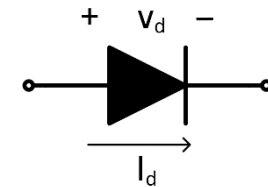
- $V_d < V_{d,on} \approx 700 \text{ mV}$
- $I_d = 0 \text{ A}$



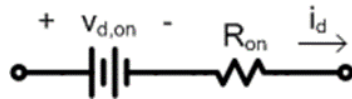
Nearly-Ideal Model with Resistance

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- Add resistance to the nearly-ideal model
 - ▣ Account for real parasitic resistance, or
 - ▣ Provide a better fit to the diode's I-V curve

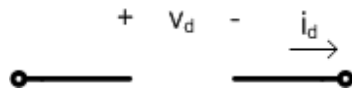


- **Forward-bias circuit:**

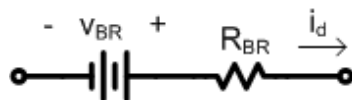


- ▣ $V_d = V_{d,on} + I_d \cdot R_{on}$

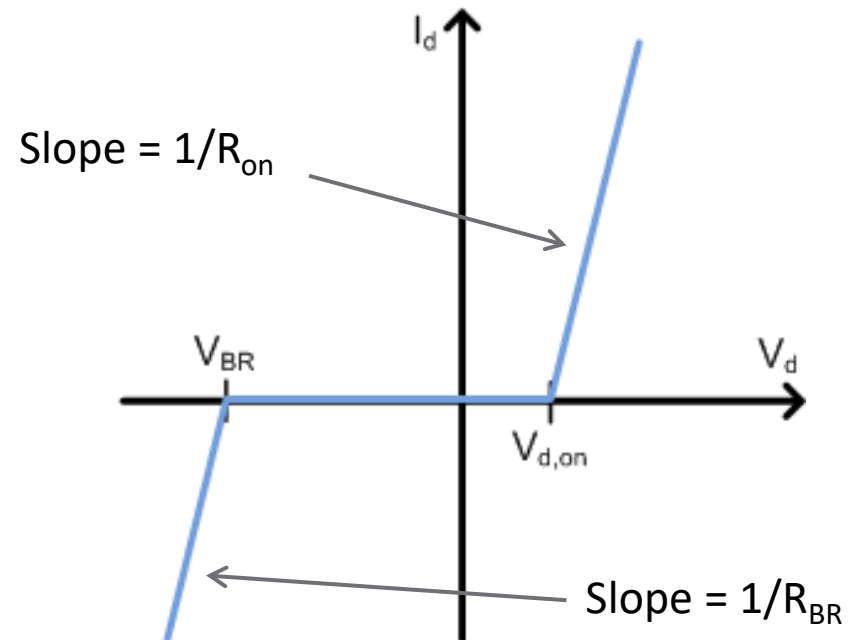
- **Reverse-bias circuit:**



- **Reverse breakdown:**



- ▣ $V_d = -V_{BR} + I_d \cdot R_{BR}$



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Load-Line Analysis

Diode Circuit Analysis

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- Analyze the circuit to find the diode operating point: V_d and I_d
- Apply KVL around the circuit

$$V_s - I_d R - V_d = 0 \quad (1)$$

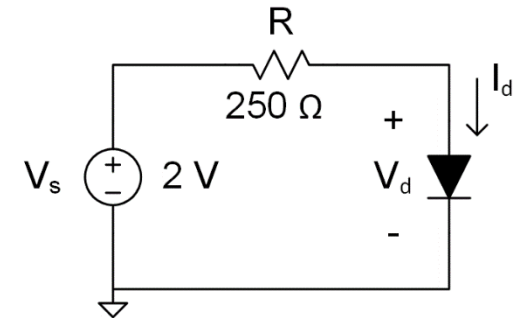
- One equation with two unknowns: V_d and I_d
 - ▣ Shockley equation give I_d in terms of V_d

$$I_d = I_s \left(e^{\frac{V_d}{V_{th}}} - 1 \right) \approx I_s e^{\frac{V_d}{V_{th}}} \quad (2)$$

- Substituting (2) into (1) yields a **transcendental equation**

$$I_s e^{\frac{V_d}{V_{th}}} \cdot R + V_d = V_s \quad (3)$$

- ▣ Solve via iteration, or
- ▣ Solve graphically – **load-line analysis**



Diode Circuit Analysis

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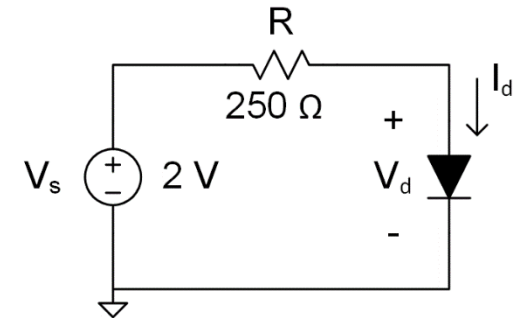
$$I_S e^{\frac{V_d}{V_{th}}} \cdot R + V_d = V_S \quad (3)$$

- Solving (3) amounts to solving the system of two equations given by (1) and (2)

$$I_d = -\frac{V_d}{R} + \frac{V_S}{R} \quad (1)$$

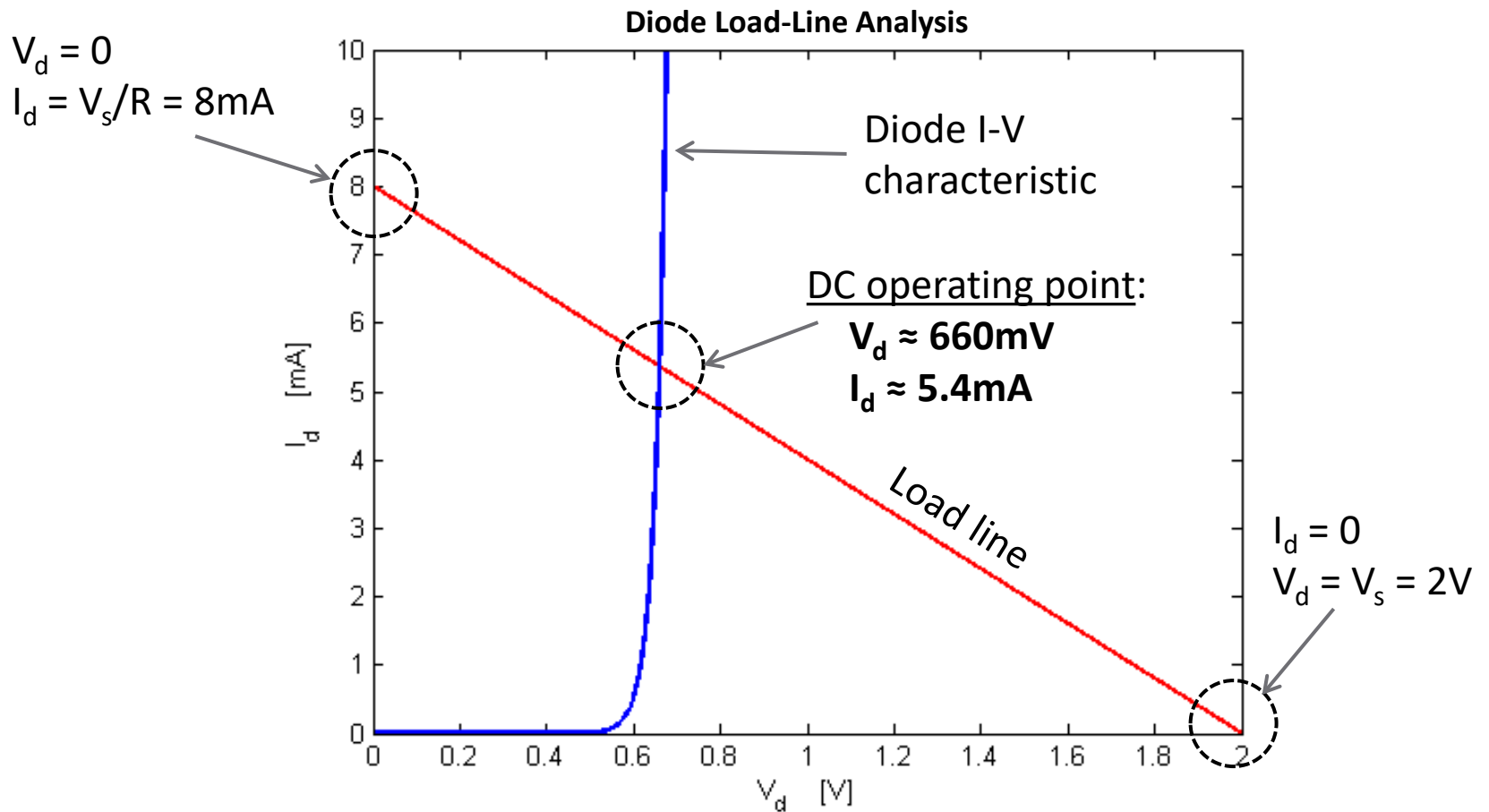
$$I_d = I_S e^{\frac{V_d}{V_{th}}} \quad (2)$$

- Equation (1) is an equation for a line – the load line
- Equation (2) is the exponential forward-biased diode characteristic
- Solution is the values of V_d and I_d that satisfy both equations
 - Point where the **two curves intersect**
 - The **DC operating point**
- Finding this solution graphically is **load-line analysis**



Load-Line Analysis

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Analysis with Simple Diode Models

Simplified Analysis – Ideal Model

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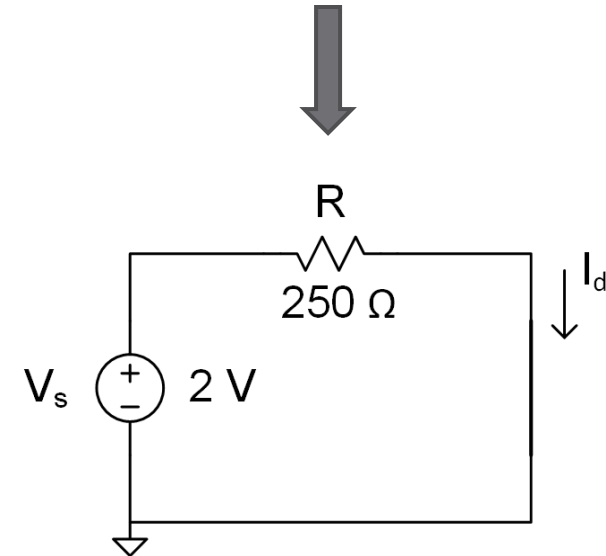
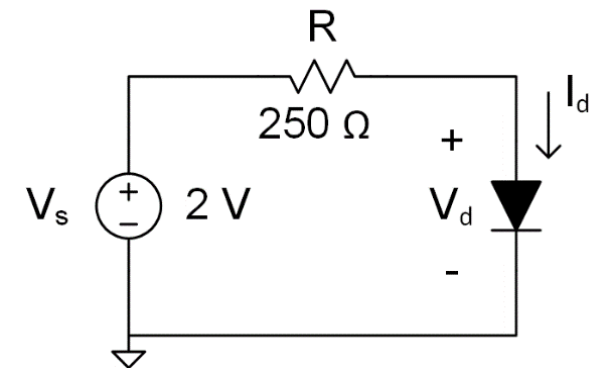
- Revisit the previous analysis using the ***ideal diode model***
- Diode is forward biased
 - ▣ Replace with a short circuit
- Diode modeled as a short, so

$$V_d = 0 V$$

- Ohm's law gives current

$$I_d = \frac{V_s}{R} = \frac{2 V}{250 \Omega} = 8 mA$$

- Current is in correct order of magnitude, but not very accurate
- Next, try the nearly-ideal model



Simplified Analysis – Nearly-Ideal Model

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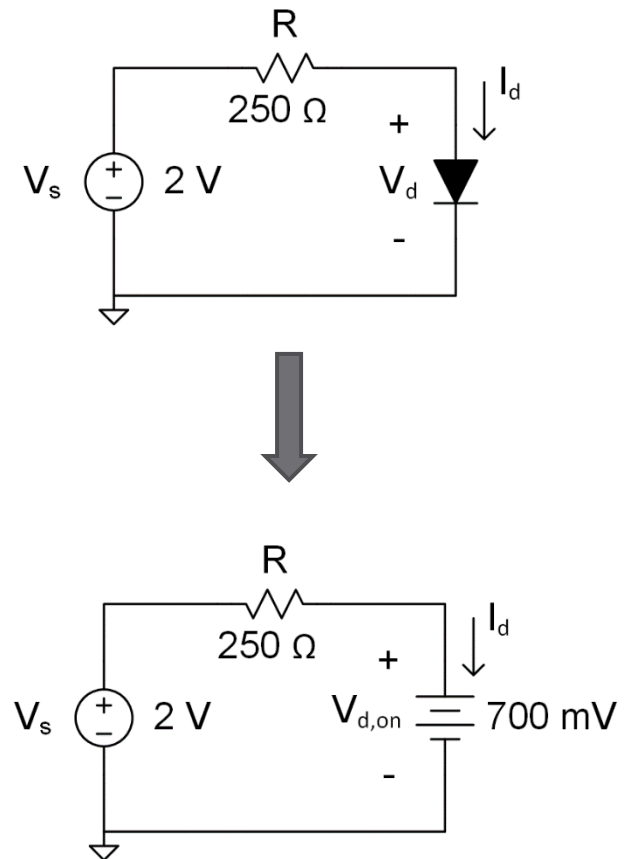
- Now, use the ***nearly-ideal model***
- Diode is forward biased
 - ▣ Replace with a voltage source
 - ▣ In practice, would have some idea of the appropriate value for $V_{d,on}$

$$V_d = V_{d,on} \approx 700 \text{ mV}$$

- Ohm's law gives current

$$I_d = \frac{V_s - V_{d,on}}{R} = \frac{2 \text{ V} - 0.7 \text{ V}}{250 \Omega} = 5.2 \text{ mA}$$

- Much more accurate result
- This is our go-to model for hand analysis



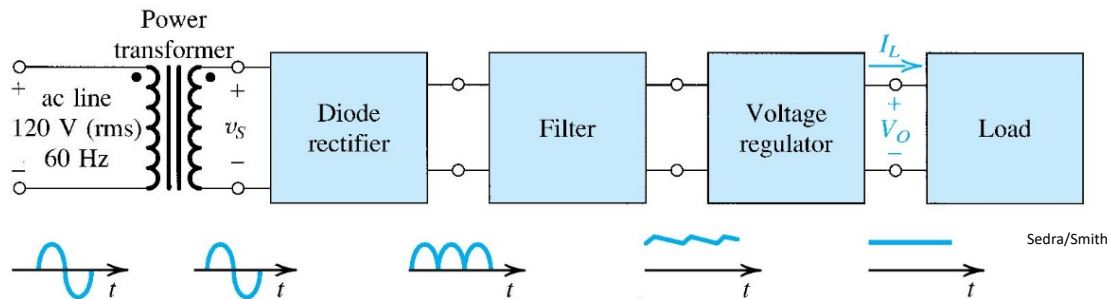
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Rectifier Circuits

Rectifier Circuits

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- **Rectifier circuits** are circuits that convert AC signals into DC signals
 - **AC-to-DC power converters:**

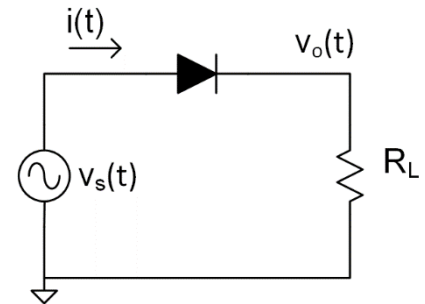


- Also used as **peak detectors**
 - AM receivers
 - Measurement instruments
- Rectifiers rely on the unidirectional nature of diodes
 - Eliminate negative voltages or make them positive

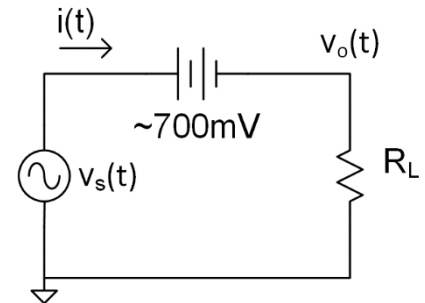
Half-Wave Rectifier

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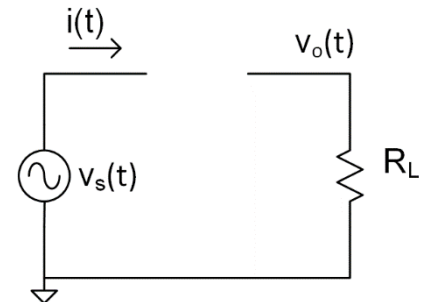
- Half-wave rectifier
 - ▣ Sinusoidal input, e.g., powerline voltage
 - ▣ Negative half-periods removed
 - ▣ Only positive voltages at rectifier output



- For $v_s(t) \geq V_{d,on} \approx 700 \text{ mV}$:
 - ▣ $i(t) > 0 \text{ A}$
 - ▣ $v_o(t) = v_s(t) - V_{d,on}$

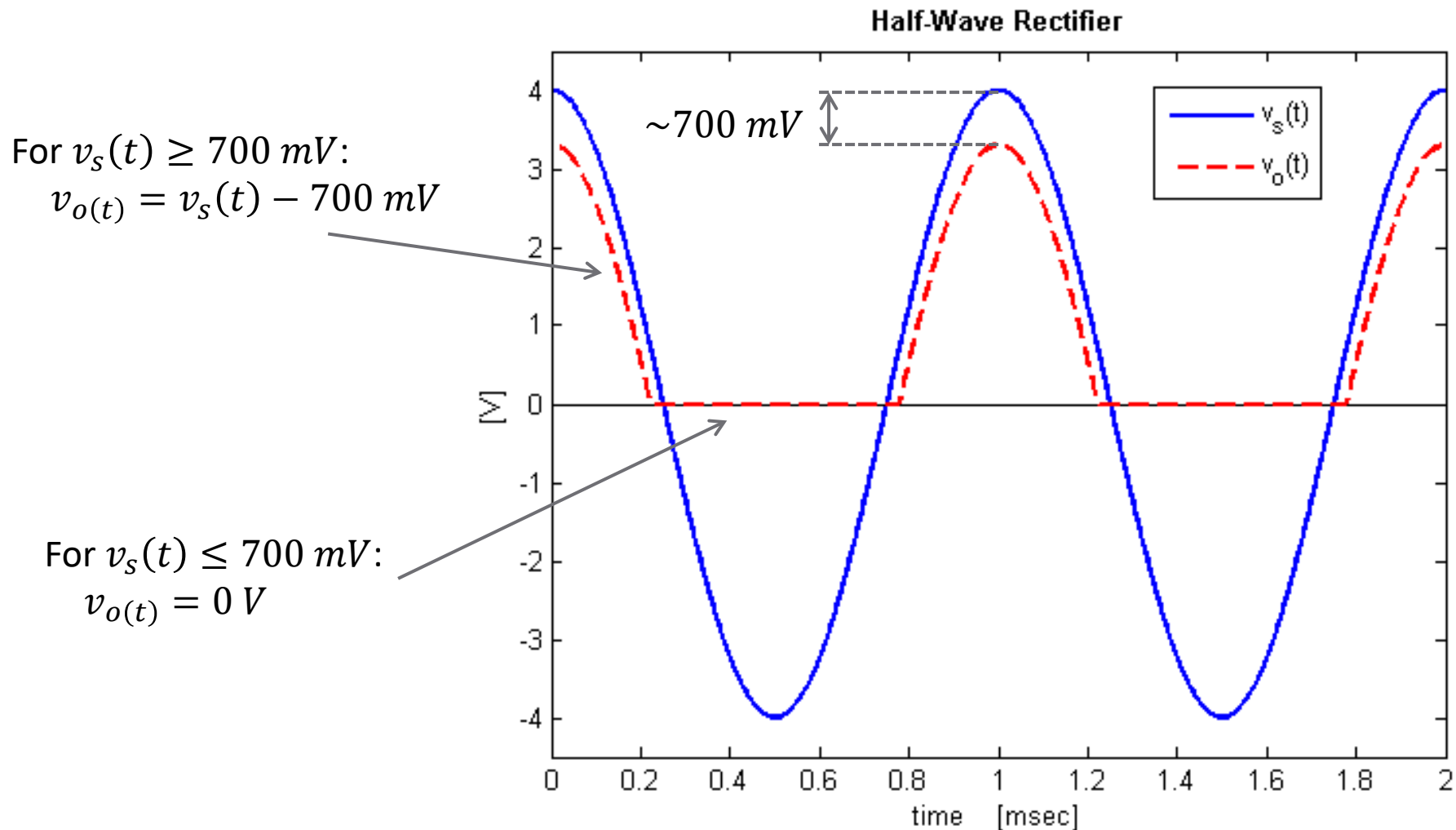


- For $v_s(t) \leq V_{d,on} \approx 700 \text{ mV}$:
 - ▣ $i(t) = 0 \text{ A}$
 - ▣ $v_o(t) = 0 \text{ V}$



Half-Wave Rectifier

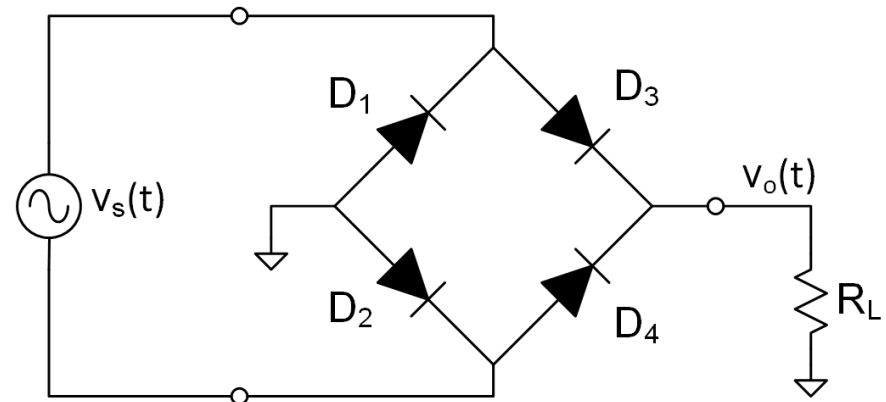
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Full-Wave Rectifier

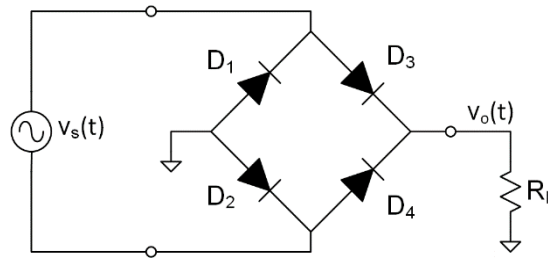
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- Typical goal of a rectifier: extract power from an AC voltage, supply it to a load as a DC voltage
- Half-wave rectification is inefficient
 - ▣ Half of the signal – and its energy – is discarded
- **Full-wave rectification** improves efficiency
 - ▣ Negative voltages are not discarded – they are made positive
- A **diode bridge** configuration
- Source must be **floating**
 - ▣ Neither side grounded
 - ▣ Typically the output of a transformer, so not a problem

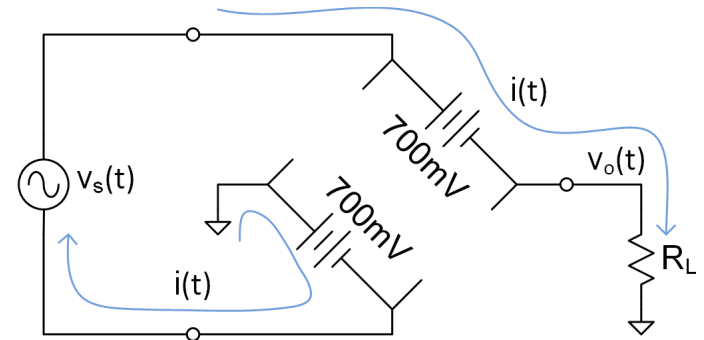


Full-Wave Rectifier

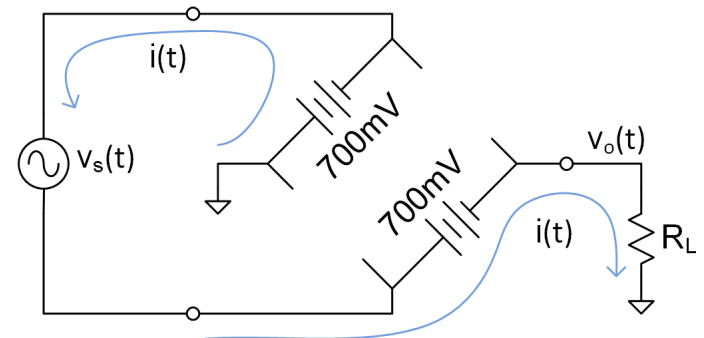
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- For $v_s(t) \geq 2 \cdot V_{d,on}$:
 - ▣ D_2 and D_3 are forward-biased
 - ▣ D_1 and D_4 are reverse-biased
 - ▣ $v_o(t) = v_s(t) - 1.4 V$

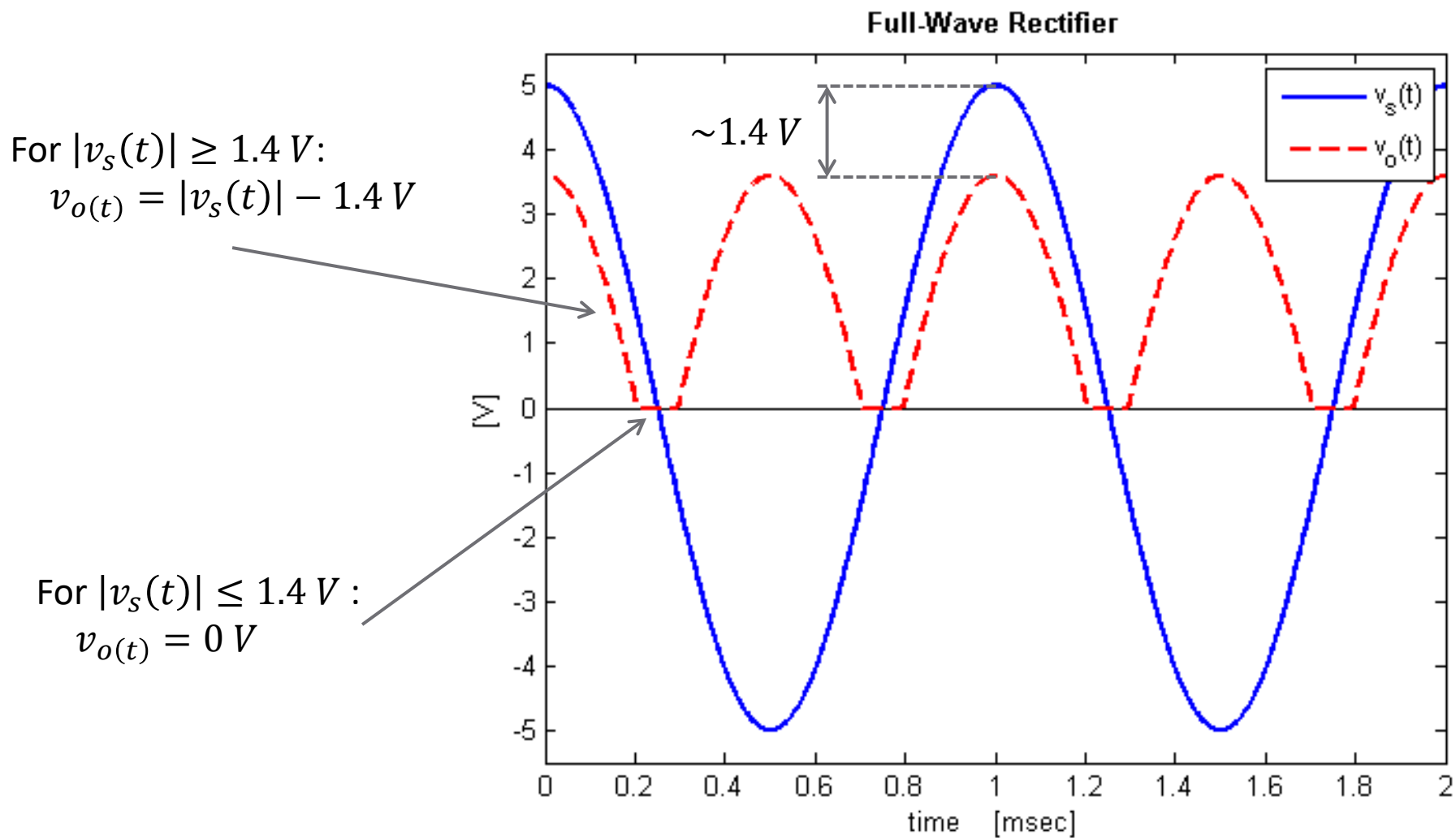


- For $v_s(t) \leq -2 \cdot V_{d,on}$:
 - ▣ D_1 and D_4 are forward-biased
 - ▣ D_2 and D_3 are reverse-biased
 - ▣ $v_o(t) = -v_s(t) - 1.4 V$



Full-Wave Rectifier

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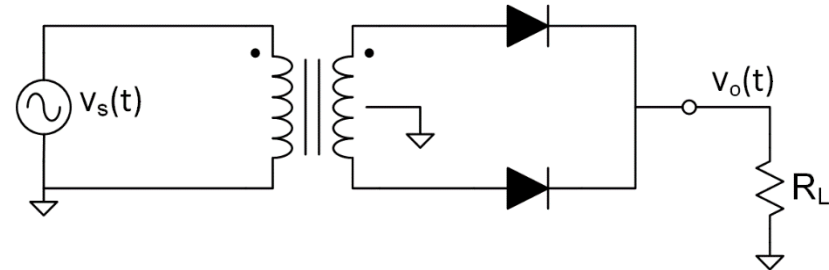
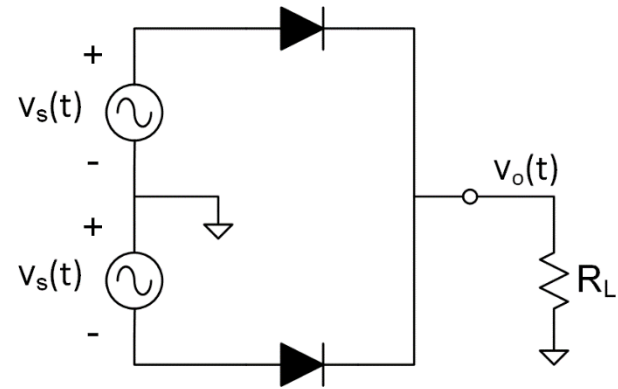
Full-Wave Rectifier – Differential Source

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- If we have a **differential source**, full-wave rectifier requires only two diodes
- Now, only a single diode drop between input and output

$$v_o(t) = |v_s(t)| - V_{d,on}$$

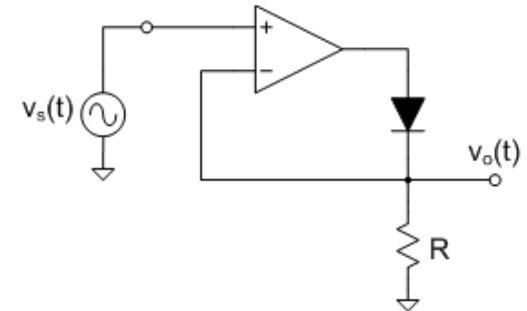
- Differential source may come from a transformer with a grounded center tap on the secondary winding



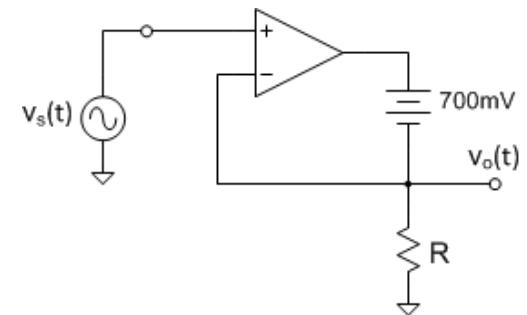
Precision Half-Wave Rectifier

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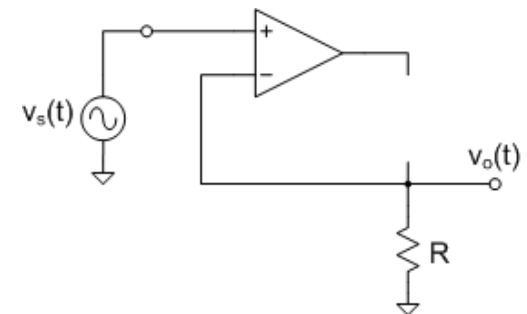
- A **precision rectifier** or **super diode** encloses a diode in an opamp feedback loop
 - ▣ Feedback forces the output equal to the input
 - ▣ Input-to-output diode drop is eliminated
 - ▣ Negative feedback only for positive inputs



- For $v_s(t) \geq 0 V$:
 - ▣ Negative feedback path exists
 - ▣ Unity-gain buffer
 - ▣ $v_o(t) = v_s(t)$



- For $v_s(t) \leq 0 V$:
 - ▣ No feedback – output saturates
 - ▣ $v_o(t)$ pulled to ground by resistor
 - ▣ $v_o(t) = 0 V$



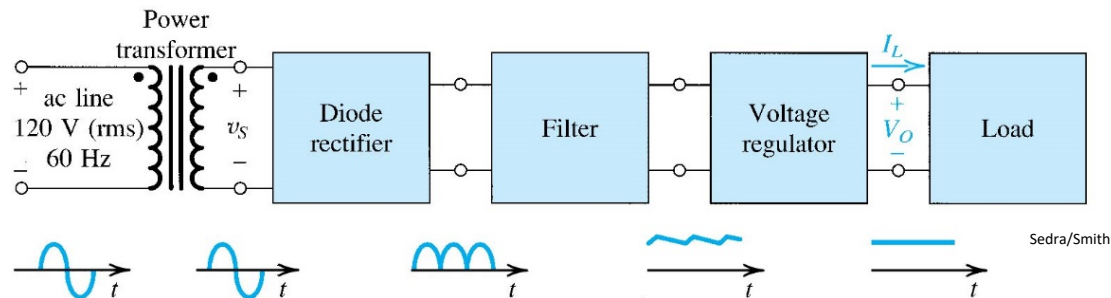
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Smoothing Capacitors & Peak Detectors

Smoothing Capacitors

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- If our goal is **AC-to-DC conversion**, rectification is only the first step
 - ▣ Rectified output is positive, but far from DC

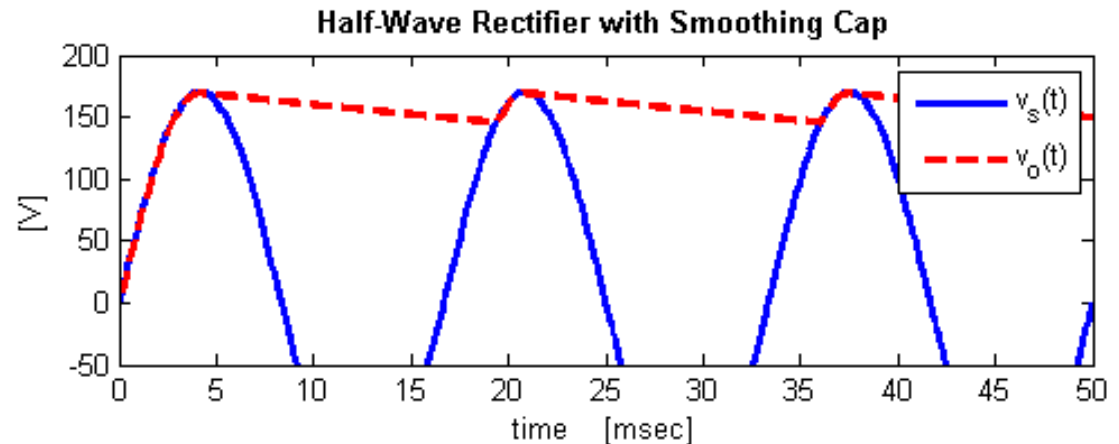
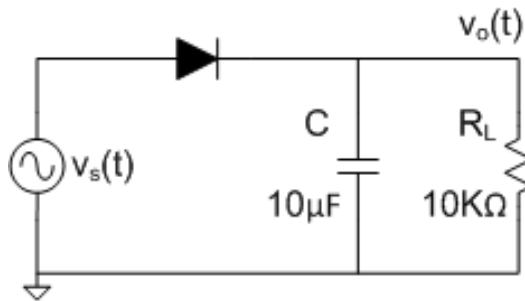


- A capacitor in parallel with the rectifier's load resistor will smooth out the output
 - ▣ A **smoothing capacitor**
 - ▣ A **low pass filter** to average out the rectified signal
- The same circuit can be used for **peak detection**
 - ▣ Demodulation of an AM radio signal, for example

Smoothing Capacitors

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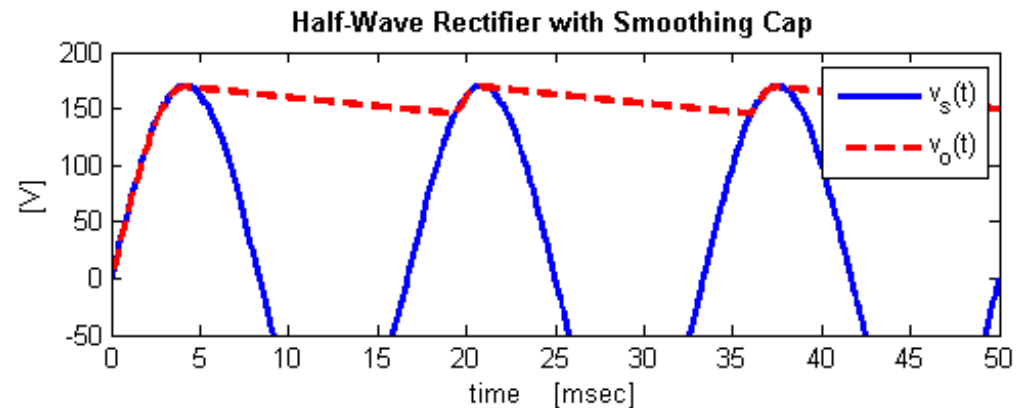
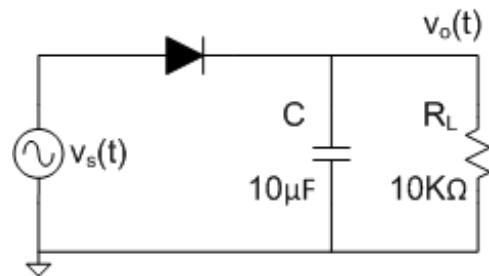
- Add a **smoothing capacitor** to half-wave rectifier
 - ▣ Input is 60 Hz powerline voltage



- When $v_o \leq v_s - V_{d,on}$:
 - ▣ Current flows through the diode
 - ▣ Capacitor charges
- When $v_s \leq v_o + V_{d,on}$:
 - ▣ Diode is off
 - ▣ Capacitor discharges at a rate determined by the RC time constant

Smoothing Capacitors

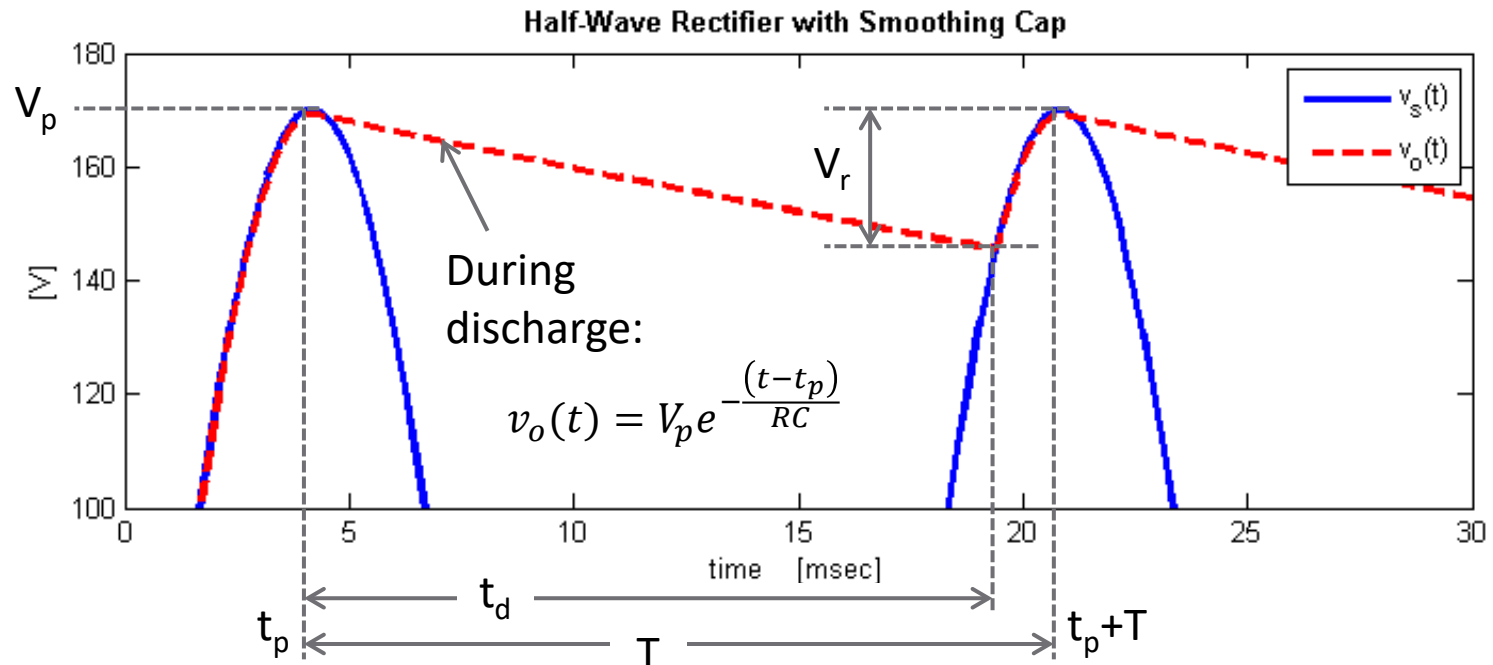
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- Capacitor helps filter or smooth the output
 - ▣ Nearly DC
 - ▣ Remaining AC component is referred to as **ripple**
- Ripple magnitude determined by how far the capacitor can discharge before charging again
 - ▣ Determined by RC time constant
- To reduce ripple
 - ▣ Increase the RC time constant
 - ▣ Larger resistor and/or capacitor

Calculating Ripple – Half Wave Rectifier

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- V_r : pk-pk ripple voltage
- V_p : peak input voltage
- t_p : time at input peak
- T : input period
- t_d : capacitor discharge time
- Diode modeled as ideal

Calculating Ripple – Half Wave Rectifier

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- V_r is the difference between the output voltage at $t = t_p$ and at $t = t_p + t_d$

$$V_r = v_o(t_p) - v_o(t_p + t_d)$$

where

$$v_o(t_p) = V_p \quad \text{and} \quad v_o(t_p + t_d) = V_p e^{-\frac{t_d}{RC}}$$

so

$$V_r = V_p - V_p e^{-\frac{t_d}{RC}} = V_p \left(1 - e^{-\frac{t_d}{RC}}\right)$$

- Problem is, we do not know t_d
 - Could calculate it, but, instead, assume the following:

$$\tau = RC \gg T$$

- From which it follows that
 - $V_r \ll V_p$
 - $t_d \approx T$
 - Discharge current is approximately constant:

$$i(t) \approx \frac{V_p}{R} \quad (t_p \leq t \leq t_p + t_d)$$

Approximating Ripple – Half Wave Rectifier

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- Recall that the change in voltage across a capacitor discharged at a constant current is

$$V = \frac{I \cdot t}{C}$$

- Using our approximations for current and discharge time gives an approximation for the ripple voltage:

$$V_r \approx \frac{V_p \cdot T}{RC} = \frac{V_p}{fRC}$$

