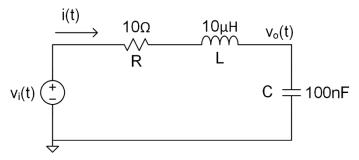
# **SECTION 1: INTRODUCTION**

ENGR 203 – Electrical Fundamentals III



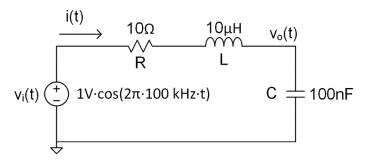
### Introduction



- In ENGR 202, we learned how to analyze circuits like this one in several ways:
  - Sinusoidal steady-state (phasor) analysis
  - **•** Frequency-response analysis
  - **D** Step-response analysis
- In ENGR 203, we will learn some new mathematical tools that we can apply to these same types of analyses
- First, we will briefly review each of the above analysis techniques



#### Sinusoidal Steady-State Analysis



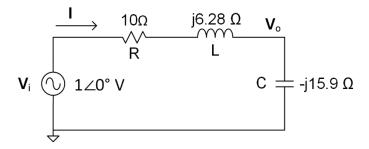
- For a sinusoidal input of a certain frequency, *phasor analysis* allows us to determine output quantities
   Also sinusoidal, at the same frequency
- □ First, convert the circuit to the phasor domain:

$$\mathbf{V}_{i} = 1 \angle 0^{\circ} V$$
  

$$Z_{L} = j\omega L = j \cdot 2\pi \cdot 100 \ kHz \cdot 10 \ \mu H = j6.28 \ \Omega$$
  

$$Z_{C} = \frac{1}{j\omega C} = -j \frac{1}{2\pi \cdot 100 \ kHz \cdot 100 \ nF} = -j15.9 \ \Omega$$

#### Sinusoidal Steady-State Analysis



 With the circuit in the phasor domain, we can apply any of the usual circuit analysis tools

■ Here, we can treat the circuit as a voltage divider

$$\mathbf{V}_{o} = \mathbf{V}_{i} \frac{-j15.9\Omega}{10 + j(6.28 - 15.9)\Omega} = 1 \angle 0^{\circ} V \frac{-j15.9\Omega}{10 - j9.62\Omega}$$
$$\mathbf{V}_{o} = 1.15 \angle -46.1^{\circ} V$$

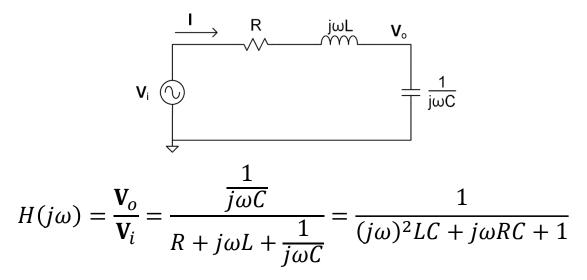
□ Finally, transform back to the time domain

 $v_o(t) = 1.15 V \cdot \cos(2\pi \cdot 100 \ kHz \cdot t - 46.1^\circ)$ 

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- Phasor analysis gets us the steady-state response of the circuit to a sinusoidal input at one specific frequency
  - The output will be at the same frequency but, in general, different amplitude and phase
- We also saw how to determine the circuit's *frequency response* function and generate its *Bode plot*
  - Response to sinusoidal input of varying frequency
  - Gain ratio of output amplitude to input amplitude
  - *Phase* phase shift from input to output

$$H(j\omega) = \frac{\mathbf{V}_o}{\mathbf{V}_i} \rightarrow |H(j\omega)|, \quad \angle H(j\omega)$$

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- Express impedances as functions of frequency
- Treat the circuit as a voltage divider

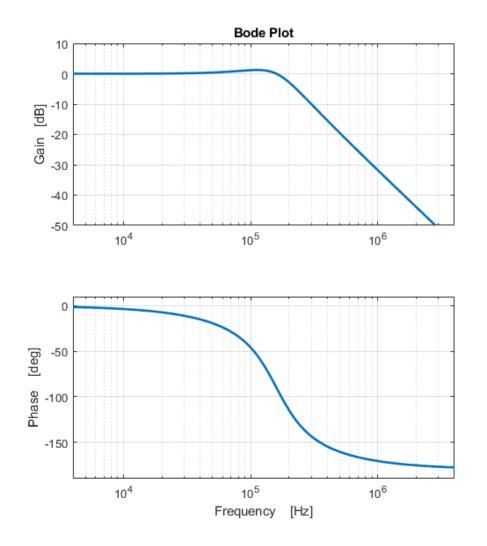


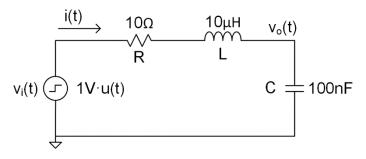
□ The resulting frequency response function is

$$H(j\omega) = \frac{\frac{1}{LC}}{(j\omega)^2 + \frac{R}{L}j\omega + \frac{1}{LC}}$$

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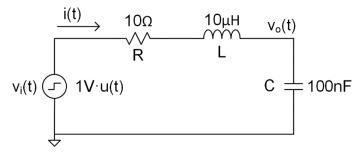
- This type of analysis is particularly useful for *filters*
  - Frequency-varying gain
  - Pass some frequencies, attenuate others





- Finally, in ENGR 202, we learned how to determine the circuit's step response
- The first step was to derive the governing differential equation
- Applying KVL around the loop yields

$$v_i(t) - i(t)R - L\frac{di}{dt} - v_o(t) = 0$$



 After eliminating i(t) and rearranging, we have the governing equation for the circuit

$$\frac{d^2 v_o}{dt^2} + \frac{R}{L} \frac{dv_o}{dt} + \frac{1}{LC} v_o(t) = \frac{1}{LC} v_i(t)$$

- □ A second-order, linear, non-homogeneous, ordinary differential equation
- Non-homogeneous, so we solve it in two parts
  - 1) Find the *complementary solution* to the homogeneous equation
  - 2) Find the *particular solution* for the step input
- General solution is the sum of the two individual solutions:

$$v_o(t) = v_{oc}(t) + v_{op}(t)$$

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#### □ For the *complementary solution*, we assume a solution of the form

$$v_{oc}(t) = e^{st}$$

where s is an unknown complex value

 Substituting this into the homogeneous equation allows us to generate the characteristic equation

$$s^2 + \frac{R}{L}s + \frac{1}{LC} = 0$$

which we can also write as

$$s^2 + 2\zeta\omega_0 s + \omega_0^2 = 0$$

where

 $\zeta$  is the damping ratio  $\omega_0$  is the natural frequency

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- The damping ratio, and the roots of the characteristic equation, determine the form of the complementary solution
  - Roots are called the circuit *poles*

#### $\Box \zeta > 1 - over-damped$

**D** Two real, distinct poles,  $s_1$  and  $s_2$ 

$$v_{oc}(t) = K_1 e^{s_1 t} + K_2 e^{s_2 t}$$

#### $\Box \ \zeta = 1 - critically-damped$

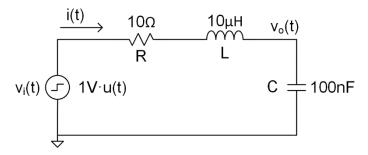
**D** Two real, identical poles,  $s_1 = s_2$ 

$$v_{oc}(t) = K_1 e^{s_1 t} + K_2 t e^{s_1 t}$$

#### $\Box \quad \zeta < 1 - Under - damped$

• Complex-conjugate pair of poles,  $s_{1,2} = -\alpha \pm j\omega_d$ 

$$v_{oc}(t) = K_1 e^{-\alpha t} \cos(\omega_d t) + K_2 e^{-\alpha t} \sin(\omega_d t)$$



The second piece of the general solution is the *particular solution* For a step input, this is simply the steady-state output

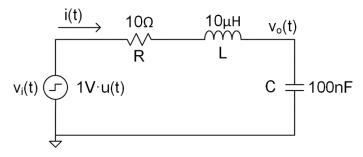
$$v_{op}(t) = v_o(t \to \infty) = 1 V$$

□ For our circuit, the characteristic equation is

$$s^{2} + \frac{R}{L}s + \frac{1}{LC} = s^{2} + 1E6s + 1E12 = s^{2} + 2\zeta\omega_{0}s + \omega_{0}^{2} = 0$$

**•** From which we find

$$\omega_0 = 1E6 \, rad/sec$$
  
 $\alpha = 500E3 \, rad/sec$   
 $\zeta = 0.5$ 



 $\Box$   $\zeta = 0.5$ , so the circuit is under-damped

- Complementary solution is a sum of damped sinusoids
- General solution has the following form

$$v_o(t) = K_1 e^{-\alpha t} \cos(\omega_d t) + K_2 e^{-\alpha t} \sin(\omega_d t) + v_o(t \to \infty)$$

The damped natural frequency is

$$\omega_d = \omega_0 \sqrt{1 - \zeta^2} = 866E3 \, rad/sec$$

General solution is

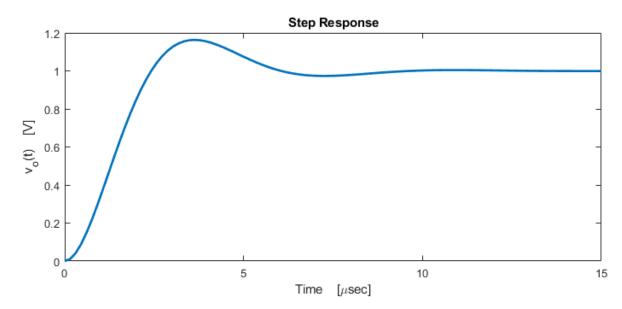
$$v_o(t) = K_1 e^{-500E3t} \cos(866E3t) + K_2 e^{-500E3t} \sin(866E3t) + 1V$$

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- □ The last step in determining the step response is to determine the unknown coefficients,  $K_1$  and  $K_2$ 
  - Apply the initial conditions:

$$v_o(0) = 0$$
 and  $\dot{v}_o(0) = 0$ 

□ The circuit's response to a 1 V step is

 $v_o(t) = -1V \, e^{-500E3t} \cos(866E3t) - 0.58V \, e^{-500E3t} \sin(866E3t) + 1V$ 

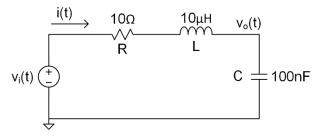


# 17 ENGR 203 Course Preview

#### **ENGR 203 Preview**

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- In this course, we will learn a new mathematical tool that will simplify and unify our circuit analyses
   The Laplace transform
- We will use the Laplace transform to:
  - Solve a circuit's governing ODE to determine its transient response to any type of input
  - Determine a circuit's time-domain response to any input without needing to derive the governing ODE
  - 3) Determine a circuit's frequency response and Bode plot
- We will now step through a preview of the application of Laplace transforms to various circuit analyses

#### Solving the Governing ODE with Laplace Transforms



- First, we will see how we will use Laplace transforms to solve a circuit's governing ODE to determine its response to any type of input
- The governing ODE for our example circuit is

$$\frac{d^2 v_o}{dt^2} + \frac{R}{L} \frac{dv_o}{dt} + \frac{1}{LC} v_o(t) = \frac{1}{LC} v_i(t)$$

Our first step will be to apply the Laplace transform to the governing equation

$$s^{2}V_{o}(s) - sv_{o}(0) - \dot{v}_{o}(0) + \frac{R}{L}[sV_{o}(s) - v_{o}(0)] + \frac{1}{LC}V_{o}(s) = \frac{1}{LC}V_{i}(s)$$

Assuming the same unit step input, and zero initial conditions, this simplifies to

$$s^{2}V_{o}(s) + \frac{R}{L}sV_{o}(s) + \frac{1}{LC}V_{o}(s) = \frac{1}{LC}\frac{1}{s}$$

K. Webb

$$s^{2}V_{o}(s) + \frac{R}{L}sV_{o}(s) + \frac{1}{LC}V_{o}(s) = \frac{1}{LC}\frac{1}{s}$$

Our differential equation has transformed to an algebraic equation

• Solve algebraically for the Laplace transform of the output,  $V_o(s)$ 

$$V_{o}(s) \left[ s^{2} + \frac{R}{L}s + \frac{1}{LC} \right] = \frac{1}{LC}\frac{1}{s}$$
$$V_{o}(s) = \frac{\frac{1}{LC}}{s\left(s^{2} + \frac{R}{L}s + \frac{1}{LC}\right)}$$
$$V_{o}(s) = \frac{1E12}{s(s^{2} + 1E6s + 1E12)}$$

K. Webb

$$V_o(s) = \frac{1E12}{s(s^2 + 1E6s + 1E12)}$$

This is the Laplace transform of the thing we are looking for, the output voltage,  $v_o(t)$ 

$$V_o(s) = \mathcal{L}\{v_o(t)\}$$

□ To get  $v_o(t)$ , we *inverse Laplace transform*  $V_o(s)$ 

$$v_o(t) = \mathcal{L}^{-1}\{V_o(s)\}$$

To inverse transform, we will make use of *partial fraction expansion* 

The partial fraction expansion looks like

$$V_o(s) = \frac{1E12}{s(s^2 + 1E6s + 1E12)} = \frac{r_1}{s} + \frac{r_2(s + 500E3) + r_3866E3}{(s + 500E3) + (866E3)^2}$$

 Solve for the unknown *residues*, r<sub>1</sub>, r<sub>2</sub>, and r<sub>3</sub>, giving the output voltage in the Laplace domain

$$V_o(s) = \frac{1}{s} - \frac{(s + 500E3)}{(s + 500E3) + (866E3)^2} - 0.58 \frac{866E3}{(s + 500E3) + (866E3)^2}$$

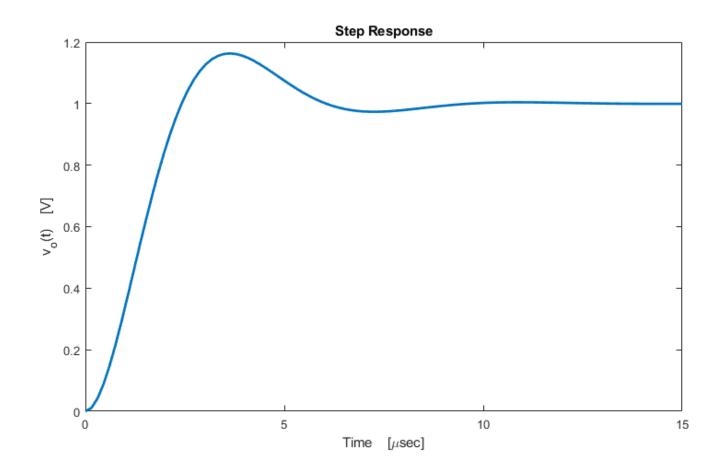
- This is now a sum of Laplace domain functions whose inverse transforms are known
  - Inverse transform by inspection

$$v_o(t) = \mathcal{L}^{-1}\{V_o(s)\}$$

 $v_o(t) = 1V - 1Ve^{-500E3t}\cos(866E3t) - 0.58Ve^{-500E3t}\sin(866E3t)$ 

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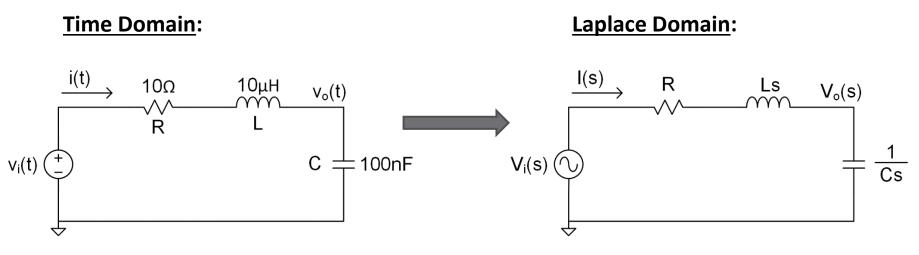
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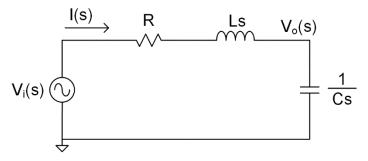


# Laplace Domain Circuit Analysis

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- Previous analysis required derivation of the governing differential equation
- Can often skip that step
  - Transform the circuit to the Laplace domain
  - Apply usual circuit analysis tools
- Similar to transforming to the phasor domain



#### Laplace Domain Circuit Analysis



 With the circuit in the Laplace domain, we can apply any of the usual circuit analysis tools

• Here, we are looking for  $V_o(s)$ , so treat it as a voltage divider

$$V_o(s) = V_i(s) \frac{\frac{1}{Cs}}{R + Ls + \frac{1}{Cs}} = V_i(s) \frac{1}{LCs^2 + RCs + 1}$$
$$V_o(s) = V_i(s) \frac{\frac{1}{LC}}{s^2 + \frac{R}{L}s + \frac{1}{LC}}$$

## Laplace Domain Circuit Analysis

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- For the given component values and a step input the output in the Laplace domain is

$$V_o(s) = \frac{1}{s} \frac{1E12}{s^2 + 1E6s + 1E12}$$

- The same result arrived at by transforming the governing ODE
  - Solve the same way: inverse Laplace transform via partial fraction expansion

$$v_o(t) = \mathcal{L}^{-1}\{V_o(s)\}$$

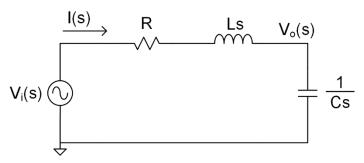
Note that this did not involve derivation of the differential equation in the time domain

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#### Laplace Domain Frequency-Response Analysis

#### Frequency Response via Laplace Transform

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- Finally, we will see how we can apply Laplace transforms to determine a circuit's *frequency response* and *Bode plot*
- First, convert the circuit to the Laplace domain



 As before, apply the usual circuit analysis techniques to determine the output

$$V_o(s) = V_i(s) \frac{\frac{1}{LC}}{s^2 + \frac{R}{L}s + \frac{1}{LC}}$$

#### Frequency Response via Laplace Transform

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$$V_o(s) = V_i(s) \frac{\frac{1}{LC}}{s^2 + \frac{R}{L}s + \frac{1}{LC}}$$

□ From here, we get the circuit's *transfer function* 

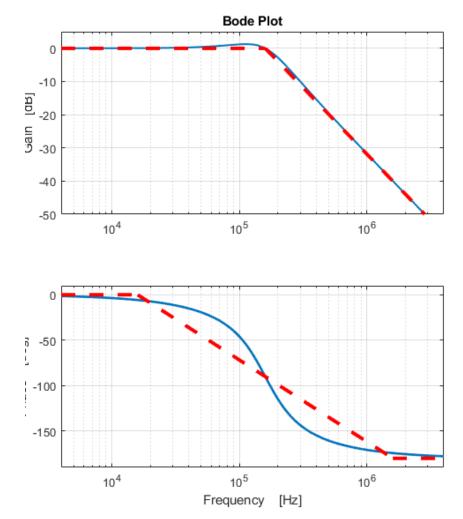
$$H(s) = \frac{V_o(s)}{V_i(s)} = \frac{\frac{1}{LC}}{s^2 + \frac{R}{L}s + \frac{1}{LC}}$$

□ From the transfer function, we get the *frequency response function* by substituting in  $j\omega$  for s

$$H(s) \xrightarrow{s \to j\omega} H(j\omega) = \frac{\frac{1}{LC}}{(j\omega)^2 + \frac{R}{L}(j\omega) + \frac{1}{LC}}$$

$$H(j\omega) = \frac{\frac{1}{LC}}{(j\omega)^2 + \frac{R}{L}(j\omega) + \frac{1}{LC}}$$

- We will learn to generate
   *Bode plot* from the
   frequency response
   function or transfer
   function
  - Hand-sketching a straightline approximation
  - Plotting in numerical tools like MATLAB or Python



# <sup>33</sup> Fourier Analysis

# Fourier Analysis

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- We will learn other mathematical tools to enable further circuit analysis
  - **•** Fourier series
  - Fourier transform
- In ENGR 202 we introduced the *frequency spectrum* 
  - Frequency content of a signal
  - How will a signal be affected by a circuit with a particular frequency response?
- Fourier series/transform allow us to mathematically determine frequency spectra
  - Spectrum of input, along with circuit's frequency response, allow us to determine spectrum of output

#### **Fourier Analysis**

